

It may “easier to approximate”
decentralized LQG problems

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Problem

$$x[n + 1] = ax[n] + u_1[n] + u_2[n] + w[n]$$

$$y_1[n] = x[n] + v_1[n]$$

$$y_2[n] = x[n] + v_2[n]$$

$$\inf_{u_1, u_2} \limsup_{N \rightarrow \infty} \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[qx^2[n] + r_1u_1^2[n] + r_2u_2^2[n]]$$

where

$$u_1[n] = f_n(y_1[0], \dots, y_1[n])$$

$$u_2[n] = g_n(y_2[0], \dots, y_2[n])$$

Problem

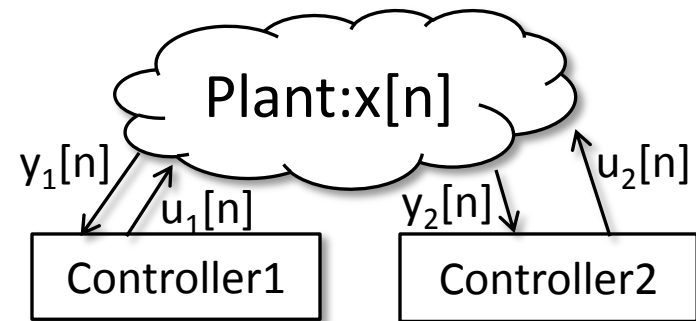
$$\begin{aligned}x[n+1] &= ax[n] + u_1[n] + u_2[n] + \boxed{w[n]} \text{ i.i.d. Gaussian} \\y_1[n] &= x[n] + \boxed{v_1[n]} \text{ i.i.d. Gaussian } \mathcal{N}(0, \sigma_{v_1}^2) \quad \mathcal{N}(0, \sigma_w^2) \\y_2[n] &= x[n] + \boxed{v_2[n]} \text{ i.i.d. Gaussian } \mathcal{N}(0, \sigma_{v_2}^2)\end{aligned}$$

$$\inf_{u_1, u_2} \limsup_{N \rightarrow \infty} \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[qx^2[n] + r_1u_1^2[n] + r_2u_2^2[n]]$$

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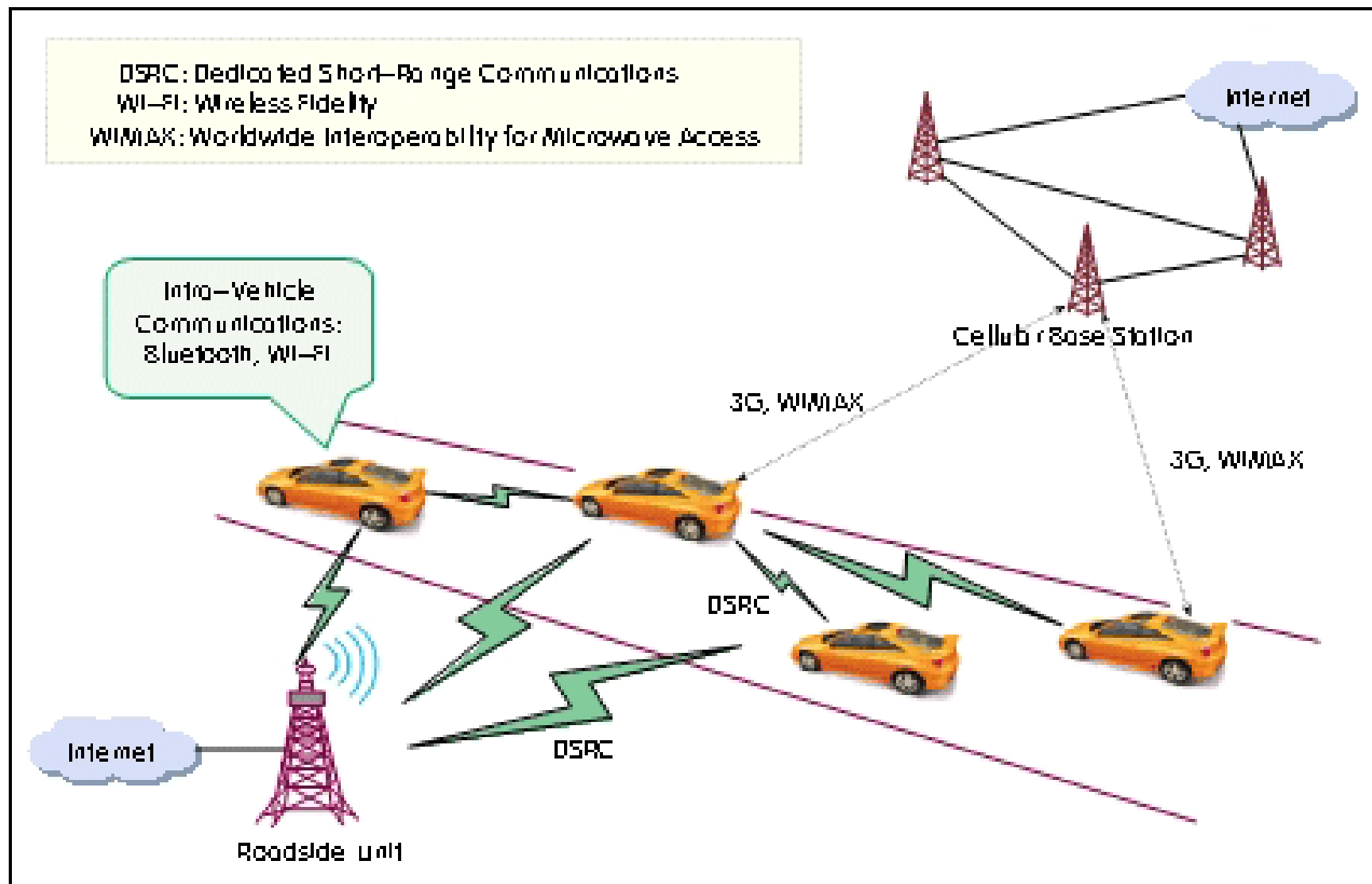
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$$u_2[n] = g_n(y_2[0], \dots, y_2[n])$$



Infinite-Horizon Average-Cost **Decentralized** Linear Quadratic
Gaussian with Scalar Plant and **Two Controllers**

Motivation: Automatic Driving System



▲ Figure 2. Wireless technologies for future vehicular communications.

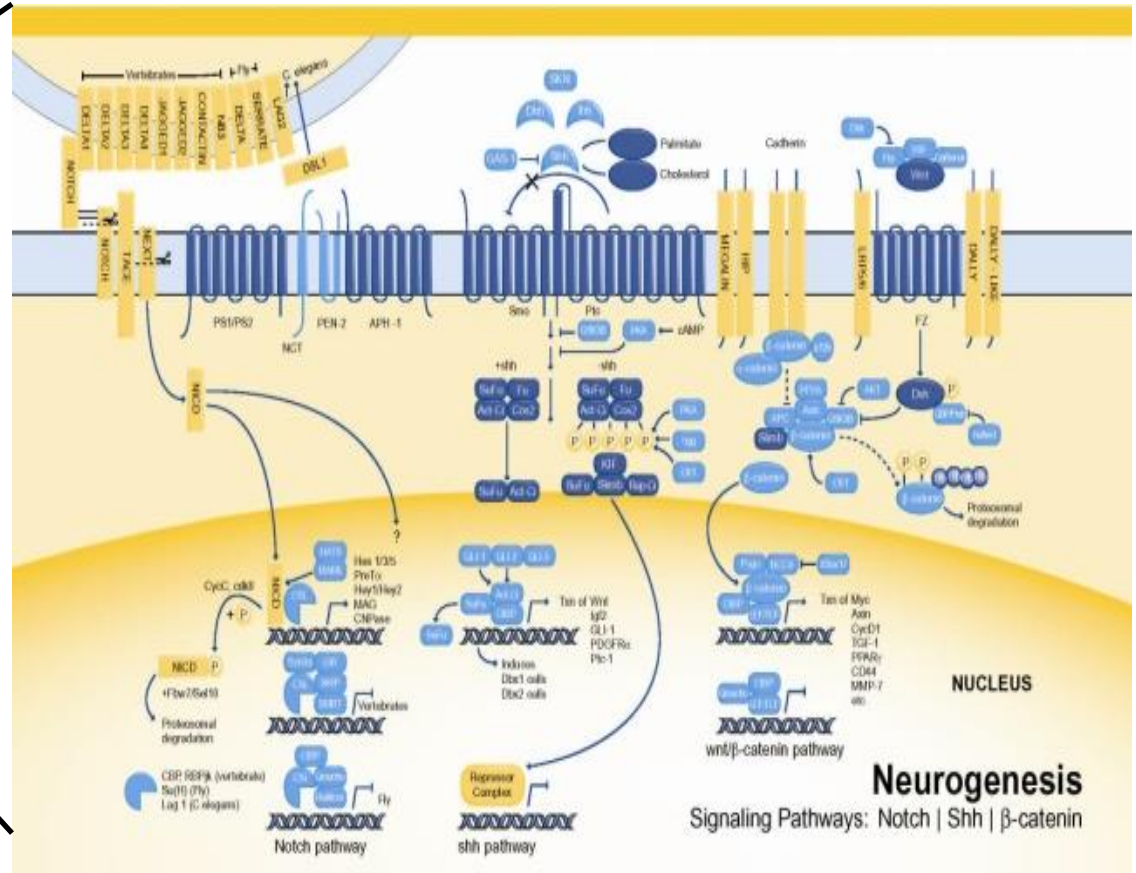
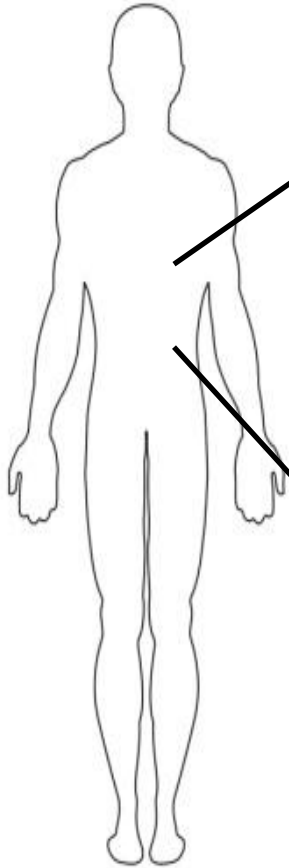
Motivation: UAV (Unmanned Aerial Vehicle)



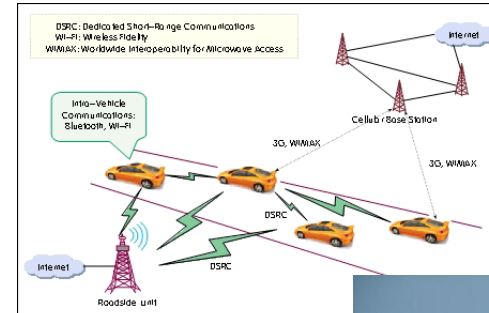
Motivation: Power Grid



Motivation: Biological System

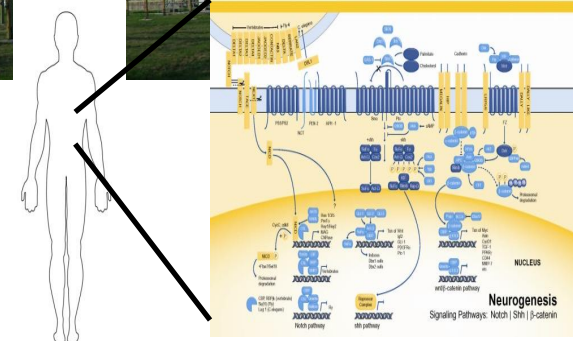


Traditional System vs Modern System



▲ Figure 2. Wireless technologies for future vehicular communication

VS



Main Difference: **Distributedness**

Decentralized LQG Problem

$$x[n + 1] = Ax[n] + B_1u_1[n] + \cdots + B_mu_m[n] + w[n]$$

$$y_1[n] = C_1x[n] + v_1[n]$$

⋮

$$y_m[n] = C_mx[n] + v_m[n]$$

$$\inf_{u_i} \limsup_{N \rightarrow \infty} \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[x^*[n]Qx[n] + \sum_{1 \leq i \leq m} u_i^*[n]R_iu_i[n]]$$

History

- **Centralized LQG Problem**

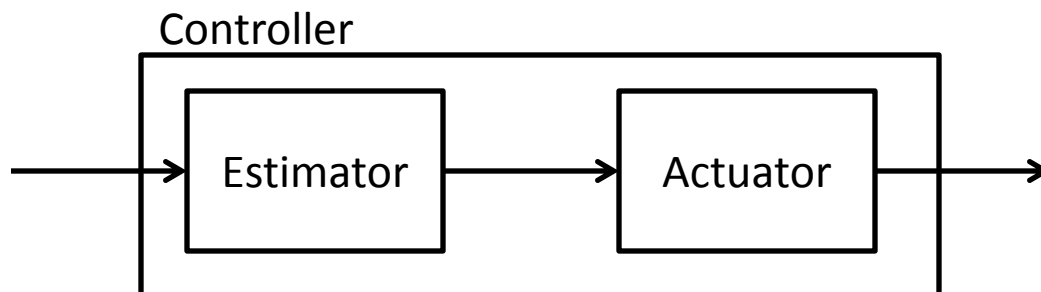
$$x[n + 1] = Ax[n] + Bu[n] + w[n]$$

$$y[n] = Cx[n] + v[n]$$

$$\inf_u \limsup_{N \rightarrow \infty} \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[x^*[n]Qx[n] + u[n]^*Ru[n]]$$

Two Lessons

- **Linear Controller** is Optimal (**Finite-Dimensional** Solution)
- **Estimation-Control** Separation



History

- **Centralized LQG Problem**

$$x[n + 1] = Ax[n] + Bu[n] + w[n]$$

$$y[n] = Cx[n] + v[n]$$

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Two Lessons

- **Linear Controller** is Optimal (**Finite-Dimensional** Solution)
- **Estimation-Control** Separation

Linear Controller is also used for **Nonlinear** Systems

Adaptive Controllers also use Estimation-Control Structure

History

- Decentralized LQG Problem

$$x[n + 1] = Ax[n] + B_1u_1[n] + \cdots + B_mu_m[n] + w[n]$$

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Negative Results

- Linear Controller is **Not** Optimal
- Estimation-Control Separation does **Not** hold

History

- Decentralized LQG Problem

$$x[n + 1] = Ax[n] + B_1u_1[n] + \cdots + B_mu_m[n] + w[n]$$

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As Optimization Problem

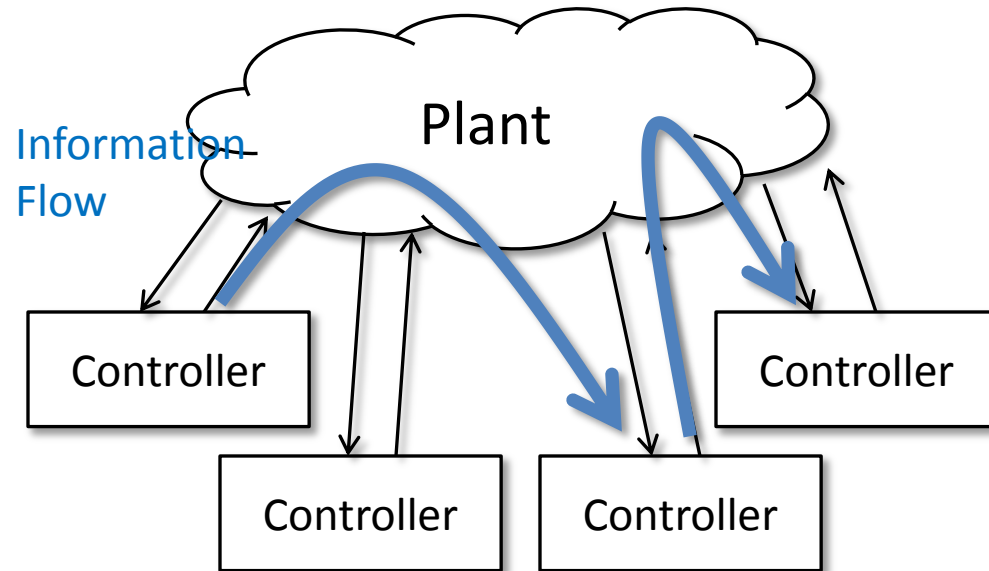
- **Infinite Dimensional** Optimization problem
- **Non-convex** problem
- **Curse of Dimensionality** from Dynamic Program

History

- Decentralized LQG Problem

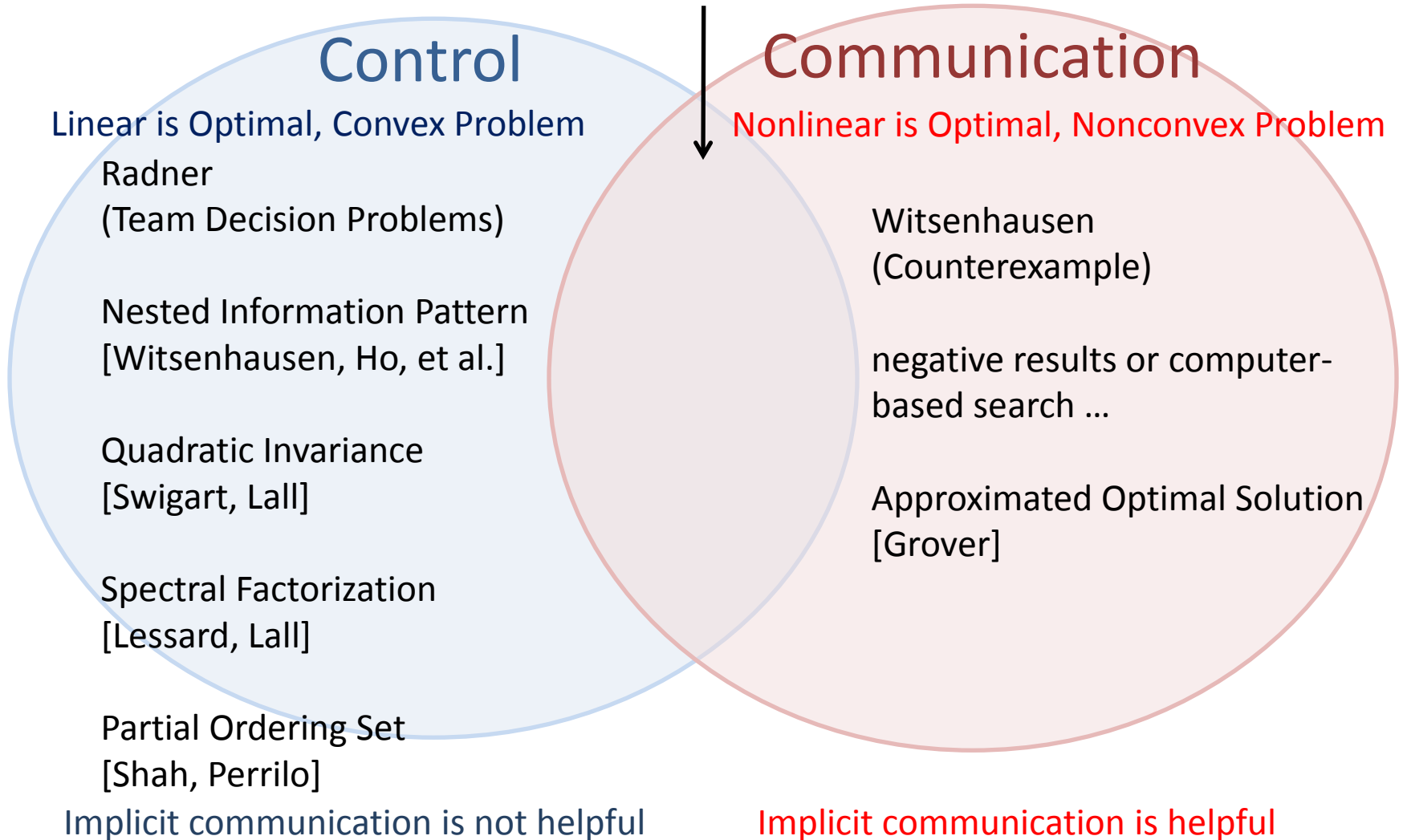
Why it is hard?

- Implicit **Communication** between Controllers



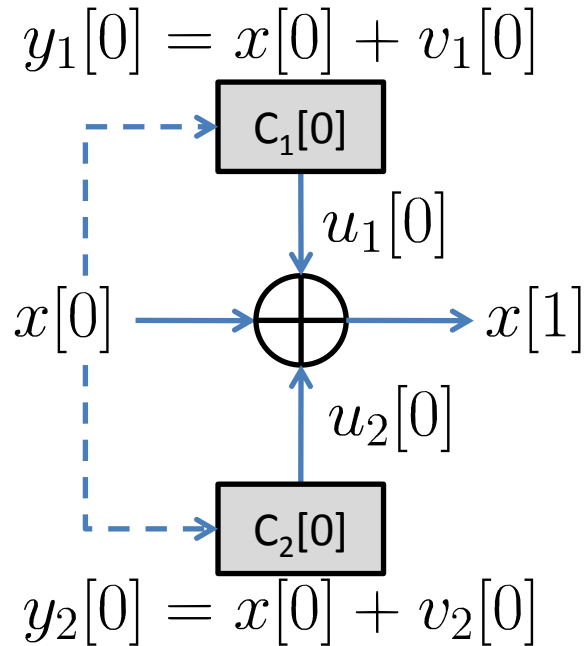
History

Decentralized LQG Problem



History

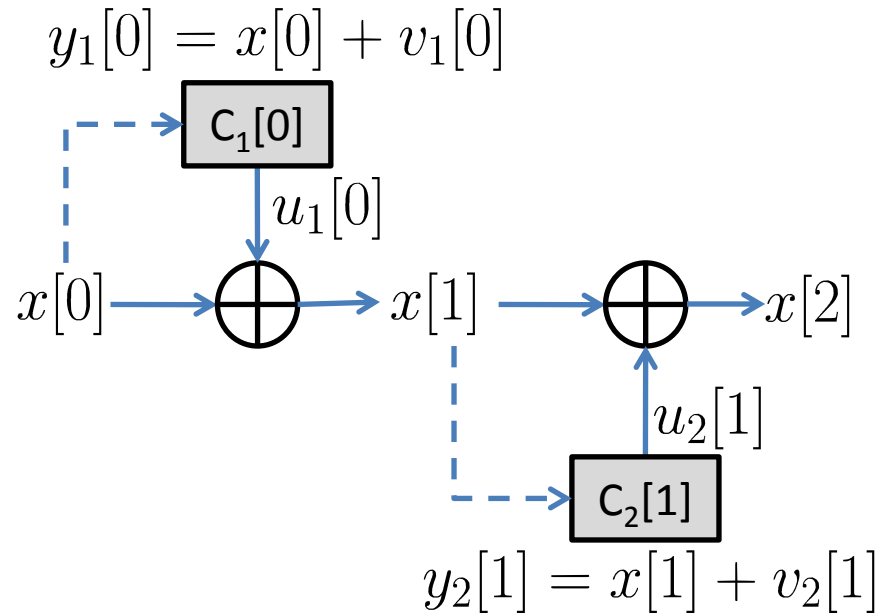
- Radner's Problem



$$\min_{u_1, u_2} \mathbb{E}[x[1]^2 + r_1 u_1^2[0] + r_2 u_2^2[0]]$$

- Linear is optimal

- Witsenhausen's Counterexample

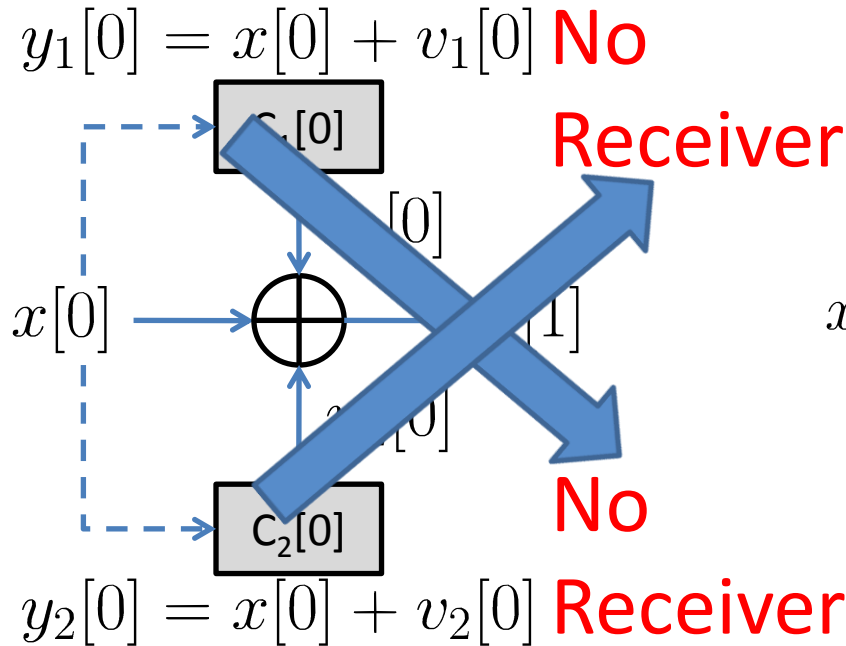


$$\min_{u_1, u_2} \mathbb{E}[x[2]^2 + r_1 u_1^2[0] + r_2 u_2^2[1]]$$

- Linear is **Not** optimal

History

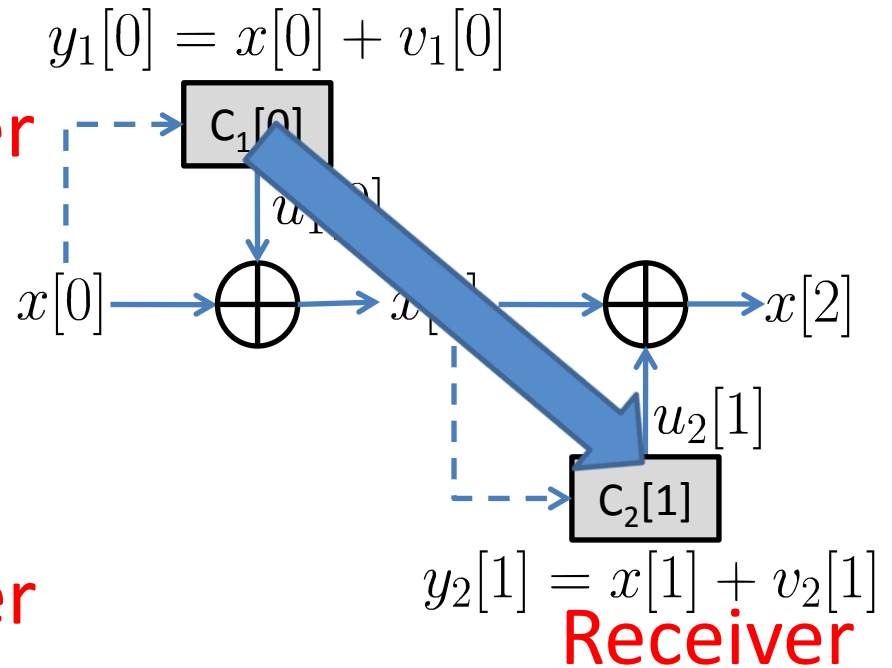
- Radner's Problem



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- Linear is **optimal**
- Implicit Communication is **Impossible**

- Witsenhausen's Counterexample



$$\min_{u_1, u_2} \mathbb{E}[x[2]^2 + r_1 u_1^2[0] + r_2 u_2^2[1]]$$

- Linear is **Not** optimal
- Implicit Communication is **possible**

Simplest Infinite-horizon LQG Problem

General Infinite-horizon Decentralized LQG Problem

$$x[n+1] = Ax[n] + B_1u_1[n] + \cdots + B_mu_m[n] + w[n]$$

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Simplest Infinite-horizon Decentralized LQG Problem:

Two Controller, Scalar Plant

$$x[n+1] = ax[n] + u_1[n] + u_2[n] + w[n]$$

$$y_1[n] = x[n] + v_1[n]$$

$$y_2[n] = x[n] + v_2[n]$$

$$\inf_{u_1, u_2} \limsup_{N \rightarrow \infty} \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[qx^2[n] + r_1u_1^2[n] + r_2u_2^2[n]]$$

Learn from Wireless Communication Theory

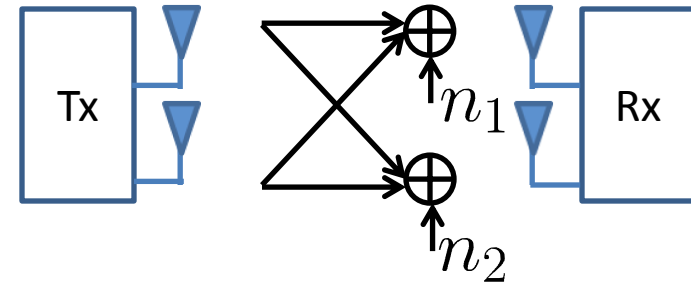
Decentralized LQG Problem

$$x[n+1] = ax[n] + u_1[n] + u_2[n] + w[n]$$

$$y_1[n] = x[n] + v_1[n]$$

$$y_2[n] = x[n] + v_2[n]$$

MIMO Communication Problem



Linear Super-position of Signals, Gaussian Disturbance

Divide Cases:

(1) Fast Dynamics
- When $|a| \geq 4$

(2) Slow Dynamics
- When $|a| < 4$

Divide Cases:

- (1) High-SNR (Signal-to-Noise Ratio)
 - d.o.f. gain (rank of signal) is important
 - Rank maximization scheme
- (2) Low-SNR (Signal-to-Noise Ratio)
 - Beam-forming gain (power of signal) is important
 - Maximum-Ratio combining scheme

Fast Dynamics Case (When $|a| \geq 4$)

- Nonlinear Controller can **infinitely** outperform Linear Controller
- We will propose **approximately optimal finite dimensional nonlinear controller**.
- Main machinery: binary deterministic model

Binary Deterministic Model (Avestimehr et al.)

- Main idea: Write a number in **binary expansion**

- Ex) Real Numbers

$$1/2 = 0.1$$

$$\pi = 11.001001\dots$$

Binary Deterministic Model (Avestimehr et al.)

- Main idea: Write a number in **binary expansion**

- Ex) Real Numbers

$$1/2 = 0.1$$

$$\pi = 11.001001\dots$$

- Ex) Random variables

$$Unif[0, 1] = 0.b_1b_2b_3\dots$$

$$Unif[0, 4] = b_1b_2.b_3b_4b_5\dots$$

where b_i are i.i.d. Bernoulli $1/2$

Binary Deterministic Model (Avestimehr et al.)

- Second Idea: **Approximate** Gaussian r.v. by Uniform r.v.

$$Unif[0, 4] = b_1 b_2 . b_3 \dots$$
$$\approx$$

$$\mathcal{N}(0, 4^2) = b_1 b_2 . b_3 \dots$$

Gaussian with zero mean and variance 4^2

Binary Deterministic Model (Avestimehr et al.)

- Third idea: **Ignore** Carry in Addition and Subtraction

Let $A \sim \mathcal{N}(0, 4^2)$ and $B \sim \mathcal{N}(0, 8^2)$

$$A = a_1 a_2 . a_3 \dots$$

$$\begin{array}{r} + B = b_1 b_2 b_3 . b_4 \dots \\ \hline \end{array}$$

$$A + B = b_1 (a_1 \oplus b_2) (a_2 \oplus b_3) . (a_3 \oplus b_4) \dots$$

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Binary Deterministic Model (Avestimehr et al.)

- Fourth idea: Multiplication and Division by a constant is **bit-shift**.

Let $A \sim \mathcal{N}(0, 4^2)$

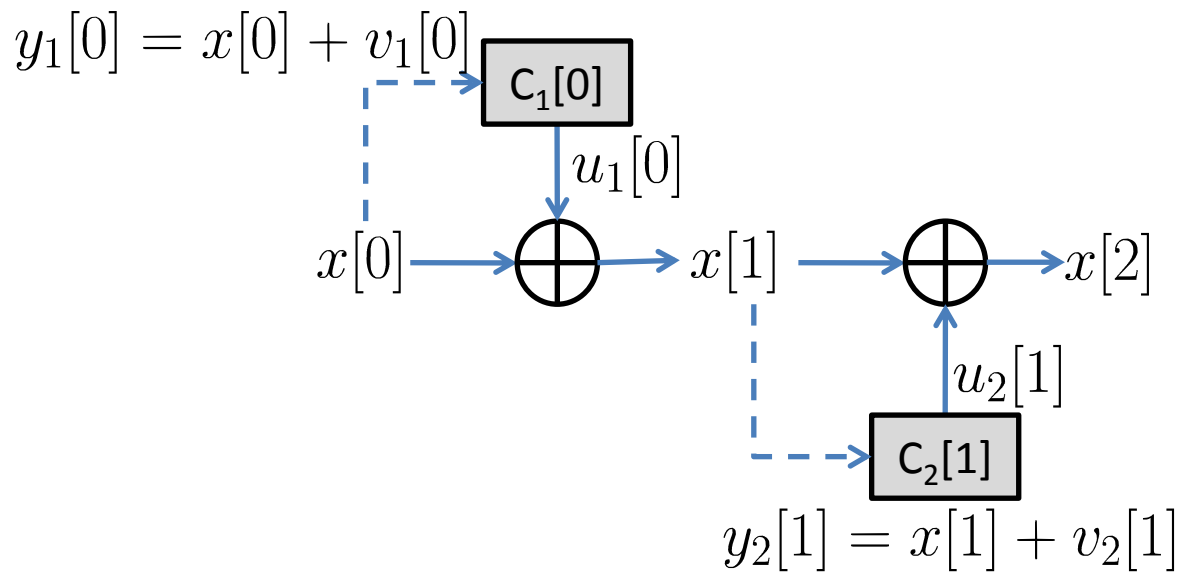
$$A = a_1a_2.a_3\dots$$

Then

$$4 \cdot A = a_1a_2a_3a_4.a_5\dots$$

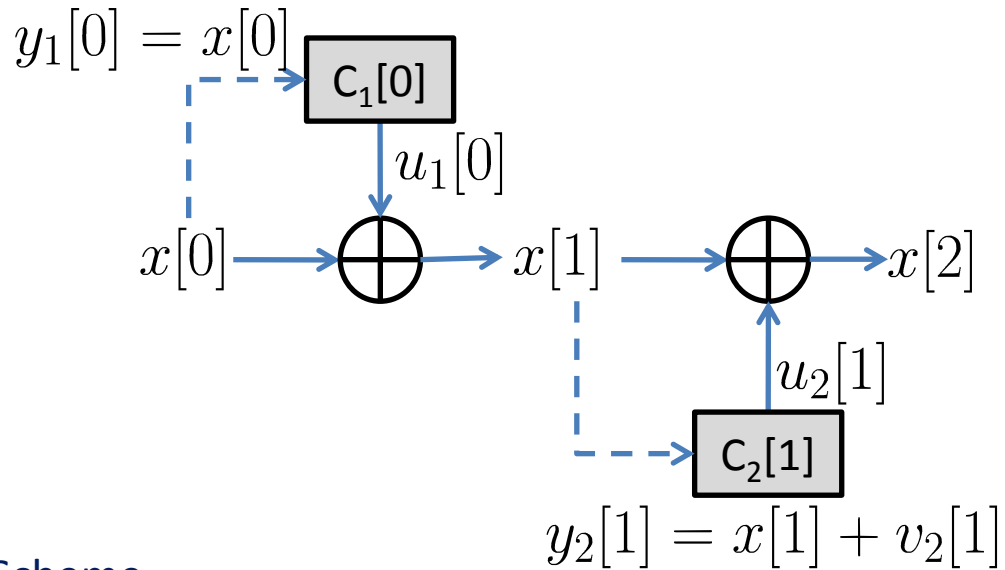
$$A/4 = 0.a_1a_2a_3a_4a_5\dots$$

Witsenhausen's Counterexample



$$\min_{u_1, u_2} \mathbb{E}[x[2]^2 + r_1 u_1^2[0] + r_2 u_2^2[1]]$$

Witsenhausen's Counterexample (Glover et al.)



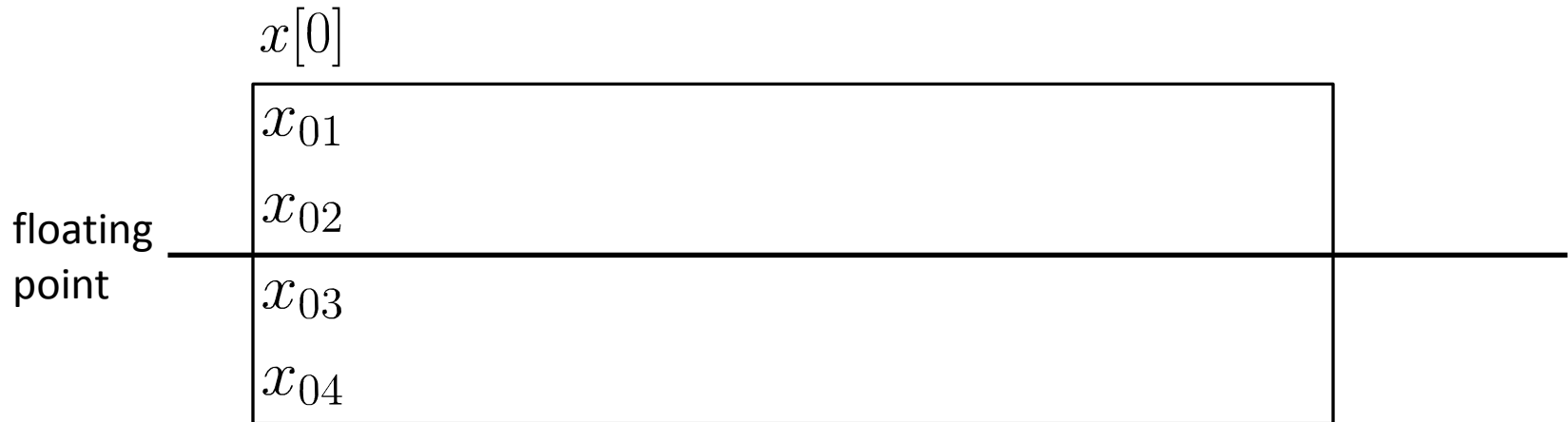
$$x[0] \sim \mathcal{N}(0, 4^2)$$

$$v_2[1] \sim \mathcal{N}(0, 1^2)$$

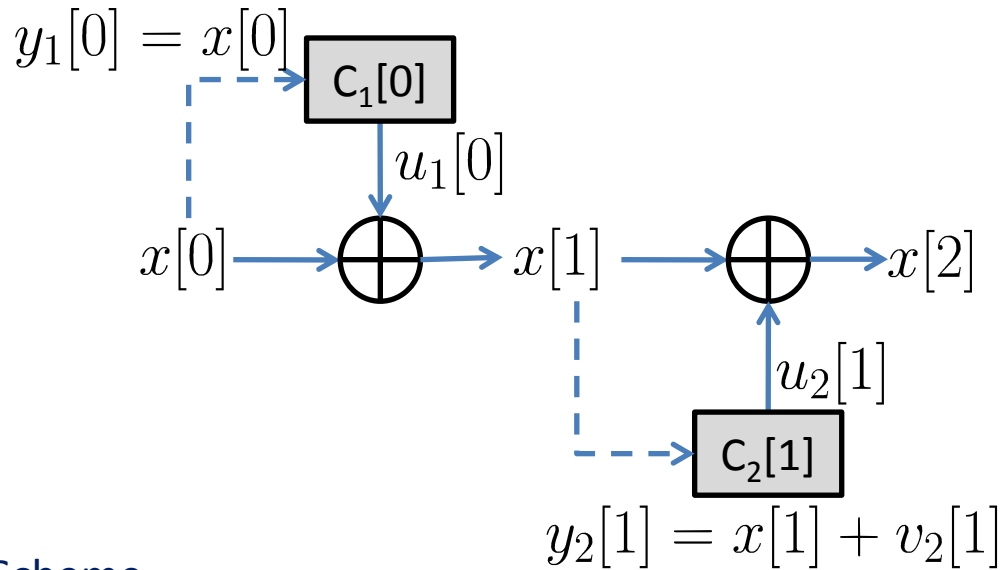
$$\min \mathbb{E}[x^2[2]]$$

$$\text{s.t. } \mathbb{E}[u_1^2[0]] \leq 1$$

Linear Scheme



Witsenhausen's Counterexample (Glover et al.)



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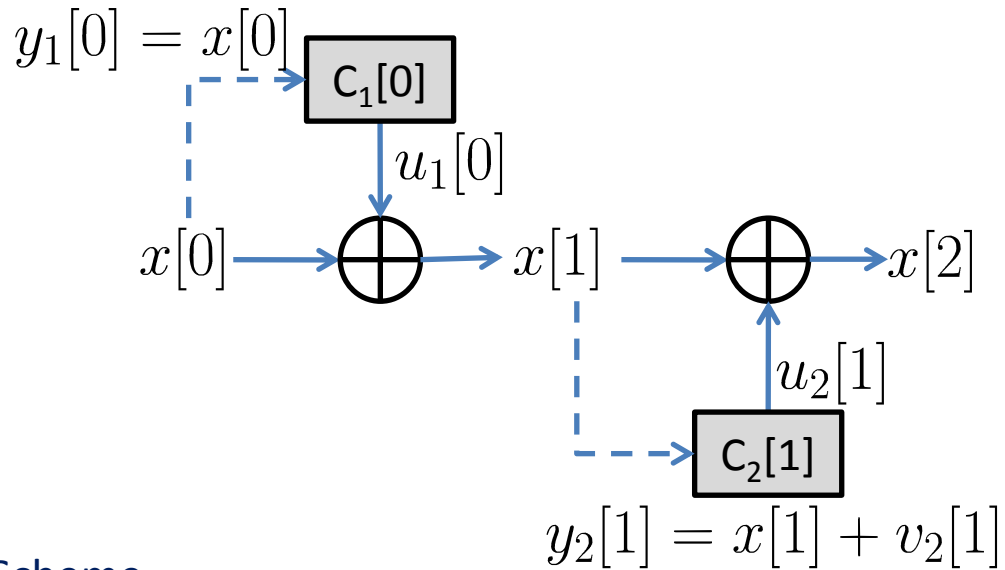
$$\min \mathbb{E}[x^2[2]]$$

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Linear Scheme

	$x[0]$	$u_1[0]$
floating point	x_{01}	0
	x_{02}	0
	x_{03}	x_{01}
	x_{04}	x_{02}

Witsenhausen's Counterexample (Glover et al.)



$$x[0] \sim \mathcal{N}(0, 4^2)$$

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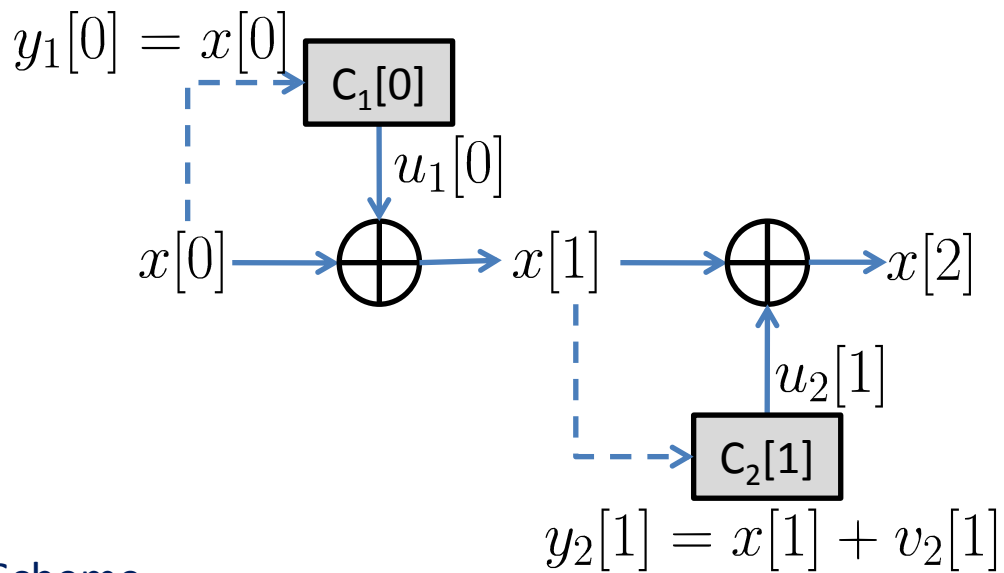
$$\text{s.t. } \mathbb{E}[u_1^2[0]] \leq 1$$

Linear Scheme

	$x[0]$	$u_1[0]$	$x[1]$
floating point	x_{01}	0	x_{11}
	x_{02}	0	x_{12}
	x_{03}	x_{01}	x_{13}
	x_{04}	x_{02}	x_{14}

$$y_2[1] = x_{11}x_{12} \cdot (x_{13} \oplus v_{21})(x_{13} \oplus v_{22})$$

Witsenhausen's Counterexample (Glover et al.)



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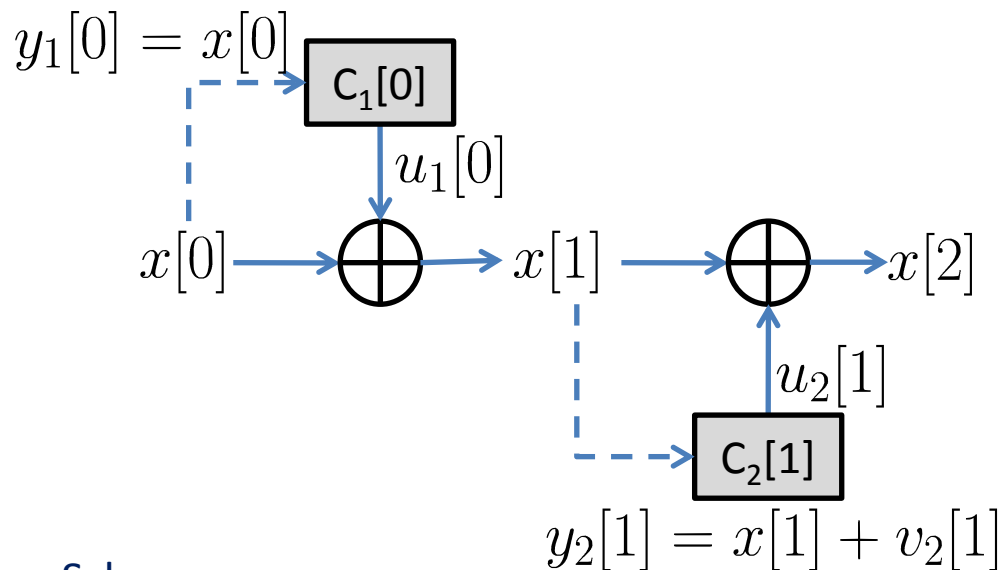
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Linear Scheme

	$x[0]$	$u_1[0]$	$x[1]$	$u_2[1]$	$x[2]$
	x_{01}	0	x_{11}	x_{11}	0
	x_{02}	0	x_{12}	x_{12}	0
floating point	x_{03}	x_{01}	x_{13}	$x_{13} \oplus v_{21}$	v_{21}
	x_{04}	x_{02}	x_{14}	$x_{14} \oplus v_{22}$	v_{22}

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Witsenhausen's Counterexample (Glover et al.)



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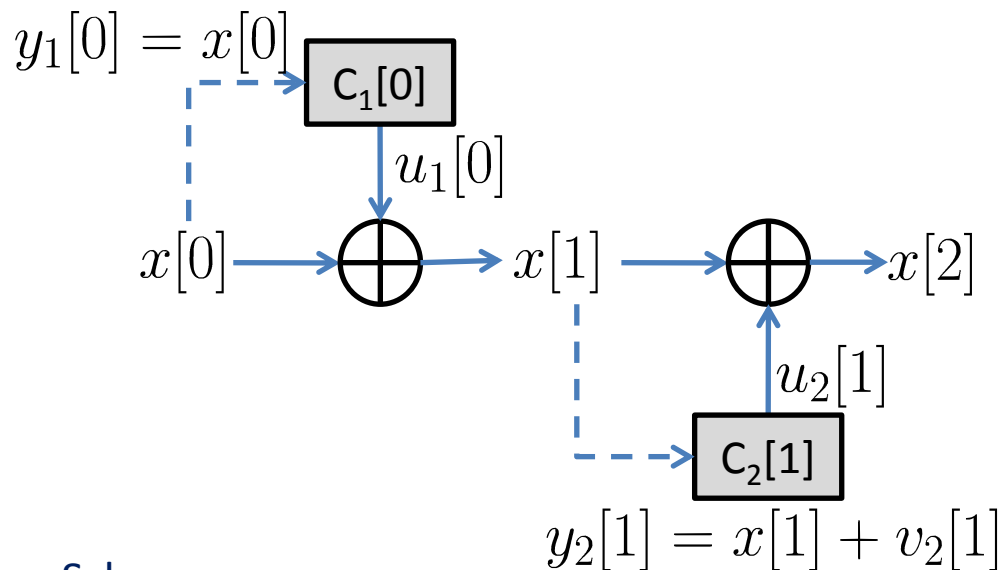
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Nonlinear Scheme

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Witsenhausen's Counterexample (Glover et al.)



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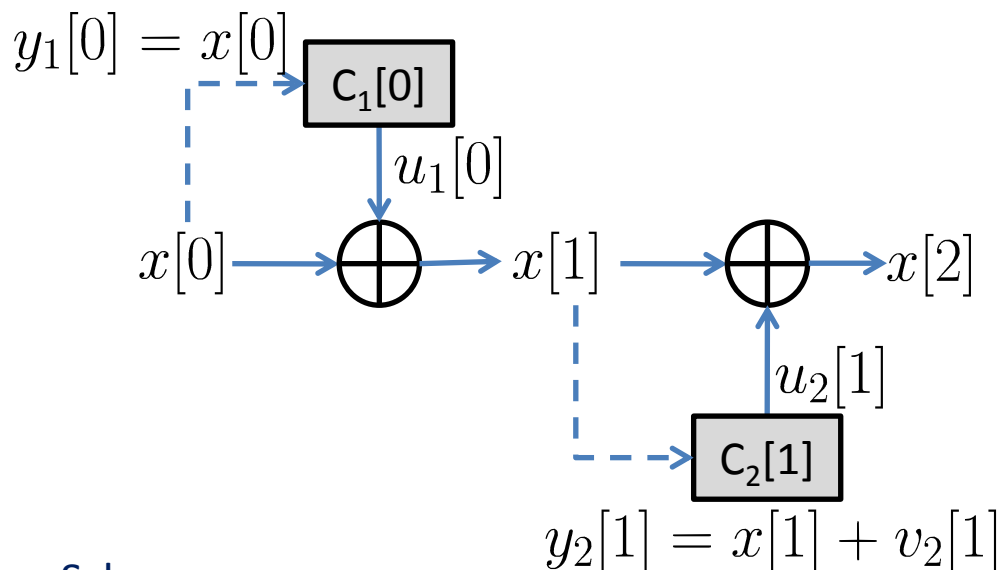
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Nonlinear Scheme

	$x[0]$	$u_1[0]$	$x[1]$
	x_{01}	0	x_{11}
	x_{02}	0	x_{12}
floating point	x_{03}	x_{03}	0
	x_{04}	x_{04}	0

$$y_2[1] = x_{11}x_{12} \cdot v_{21}v_{22}$$

Witsenhausen's Counterexample (Glover et al.)



$$x[0] \sim \mathcal{N}(0, 4^2)$$

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floating point	x_{03}	x_{03}	0	0	0
	x_{04}	x_{04}	0	0	0

$$y_2[1] = x_{11}x_{12} \cdot v_{21}v_{22}$$

Witsenhausen's Counterexample (Glover et al.)

- From deterministic model to Reals

Controller 1

$$y_1[0] = x_{01}x_{02}.x_{03}x_{04} \longrightarrow u_1[0] = 00.x_{03}x_{04}$$

Controller 2

$$y_2[1] = x_{11}x_{12}.v_{21}v_{22} \longrightarrow u_2[1] = x_{11}x_{12}.00$$

Witsenhausen's Counterexample (Glover et al.)

- From deterministic model to Reals

Controller 1

$$y_1[0] = x_{01}x_{02}.x_{03}x_{04} \longrightarrow u_1[0] = 00.x_{03}x_{04}$$

$u_1[0]$ is the remainder of $y_1[0]$ divided by 1

$$u_1[0] := R_1(y_1[0])$$

Controller 2

$$y_2[1] = x_{11}x_{12}.v_{21}v_{22} \longrightarrow u_2[1] = x_{11}x_{12}.00$$

$u_2[1]$ is the quotient of $y_2[1]$ divided by 1

$$u_2[1] := Q_1(y_2[1])$$

Linear Controller

$$x[n + 1] = 4x[n] + u_1[n] + u_2[n] + w[n]$$

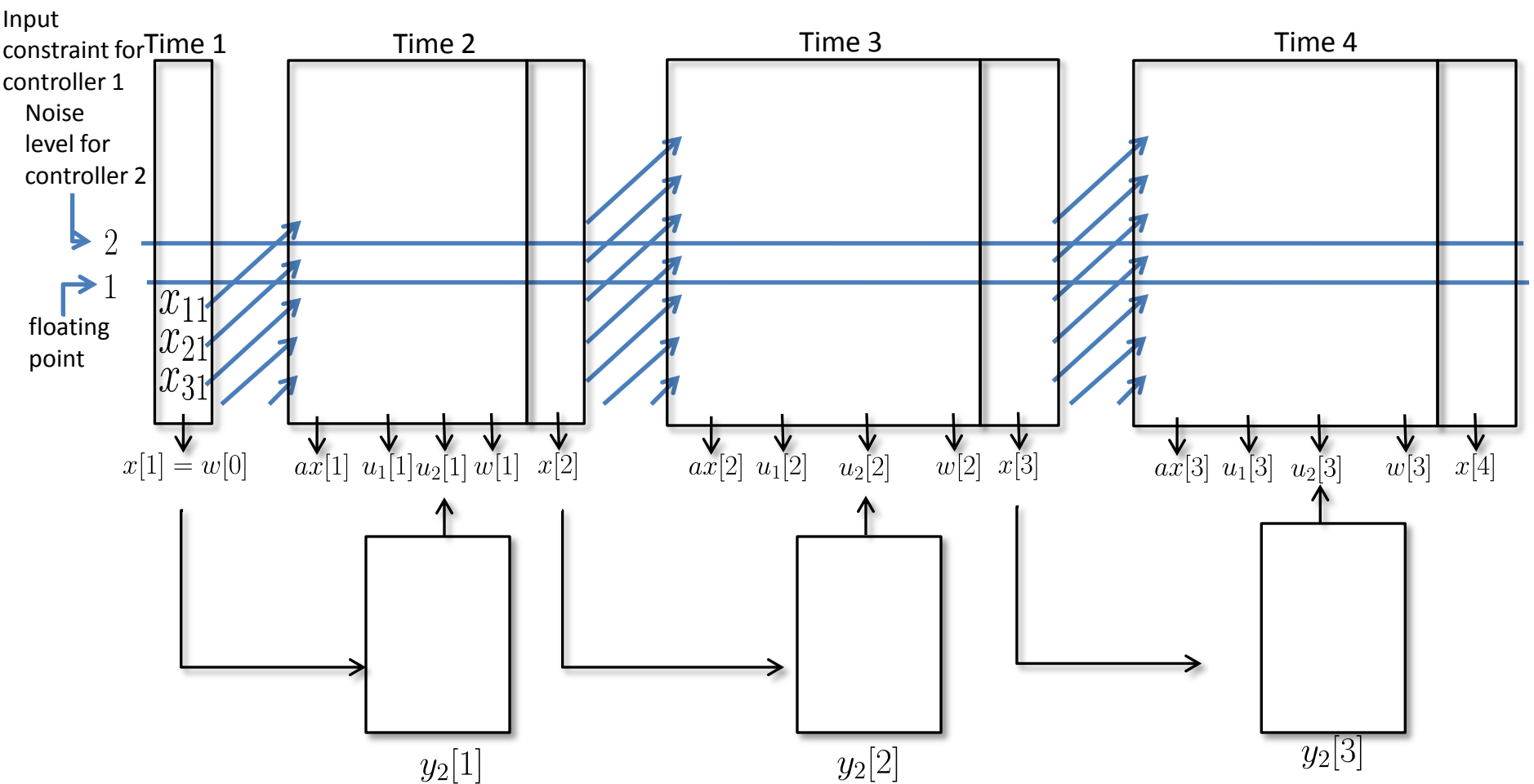
$$y_1[n] = x[n]$$

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$$w[n] \sim \mathcal{N}(0, 1)$$

$$v_2[n] \sim \mathcal{N}(0, 2^2)$$

$$\mathbb{E}[u_1^2[n]] \leq 2^2$$



Linear Controller

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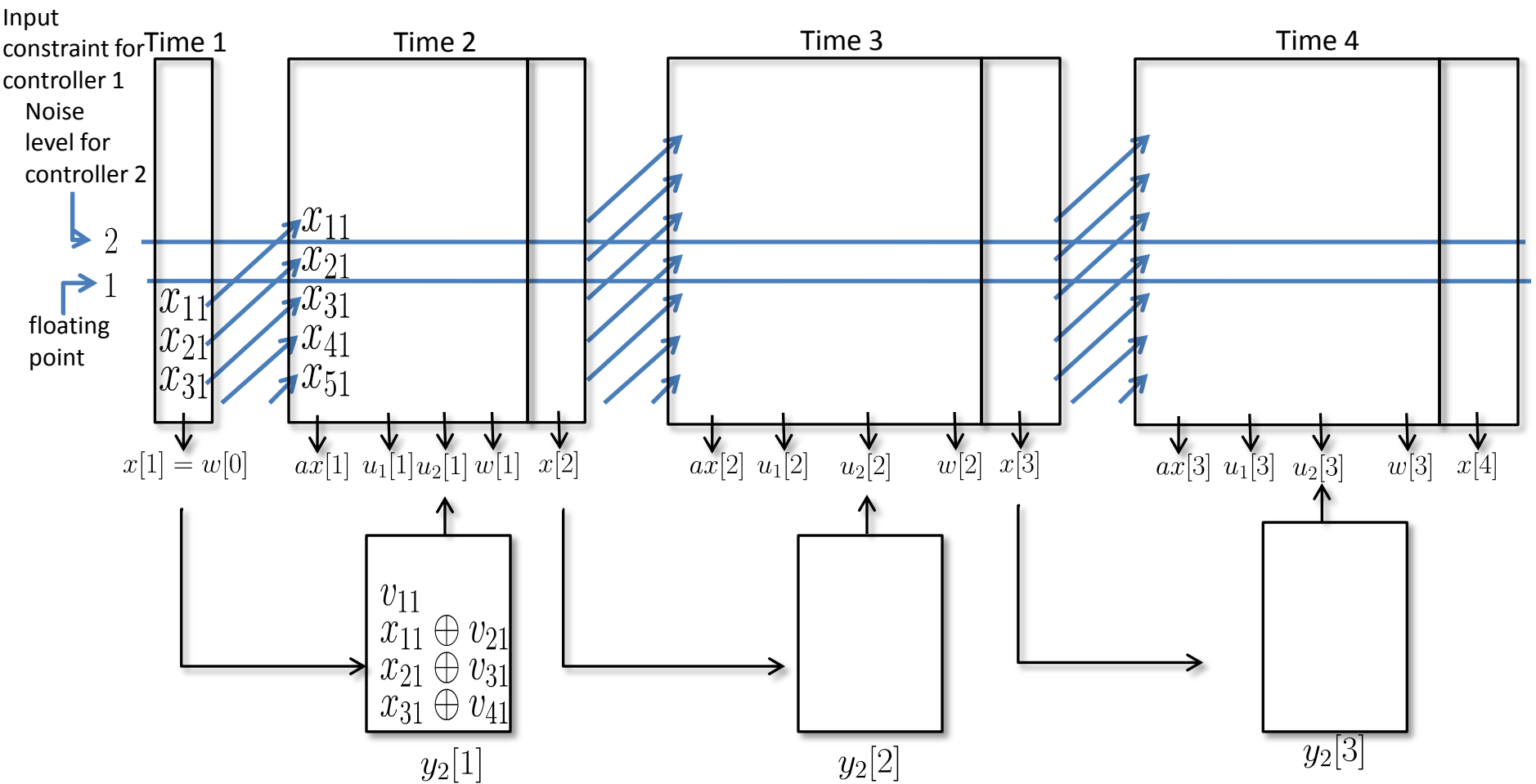
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Linear Controller

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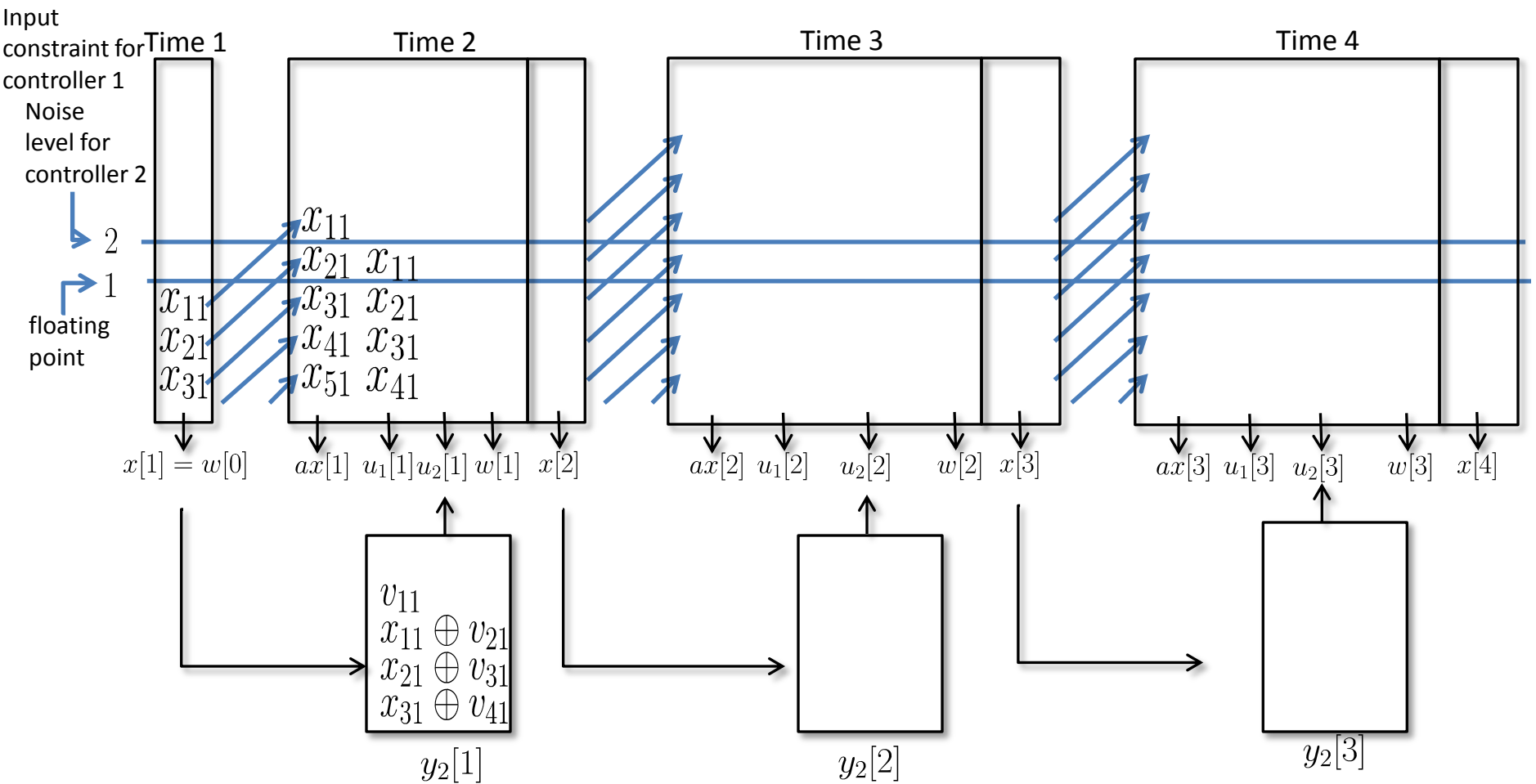
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Linear Controller

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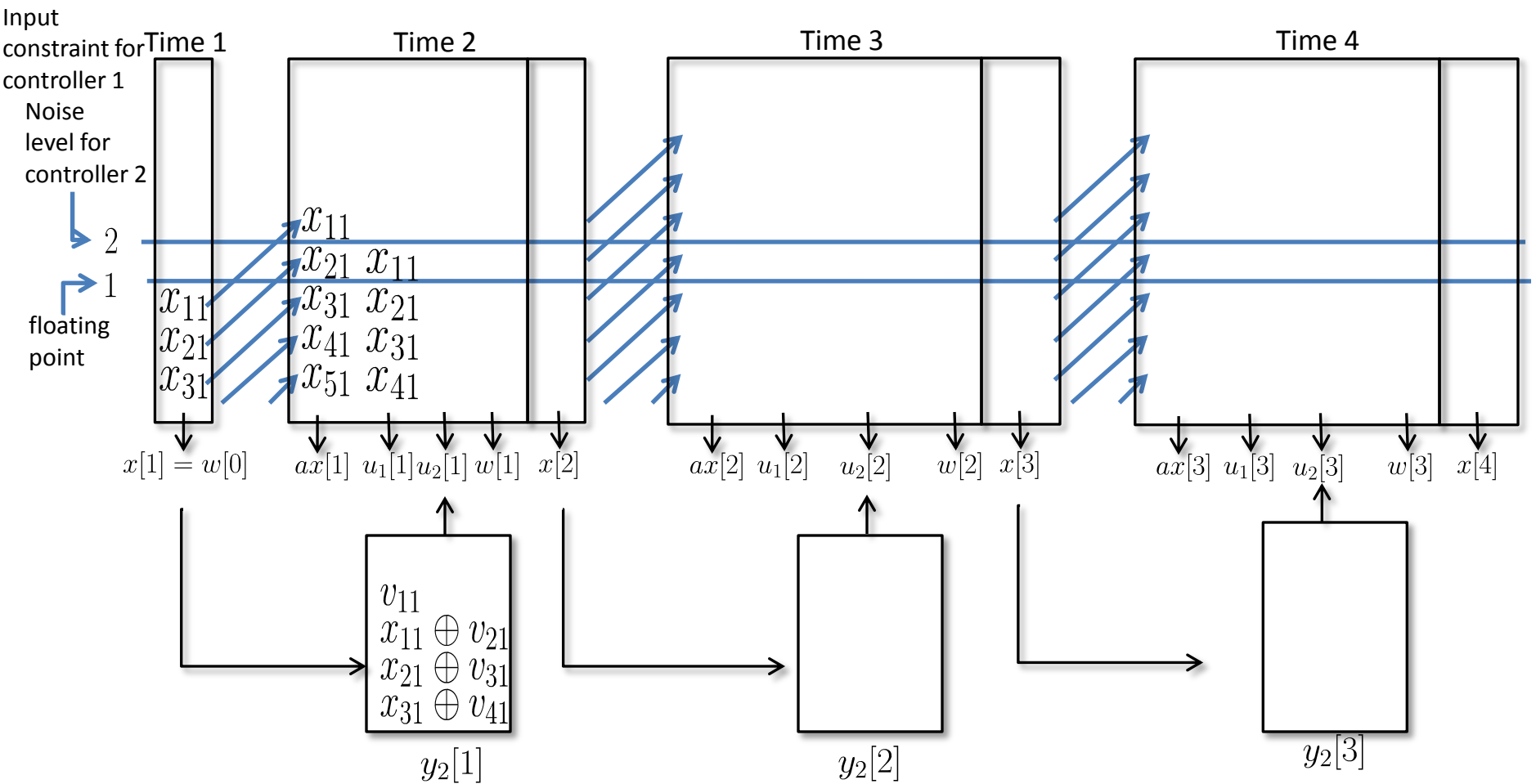
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$$\mathbb{E}[u_1^2[n]] \leq 2^2$$



Linear Controller

$$x[n + 1] = 4x[n] + u_1[n] + u_2[n] + w[n]$$

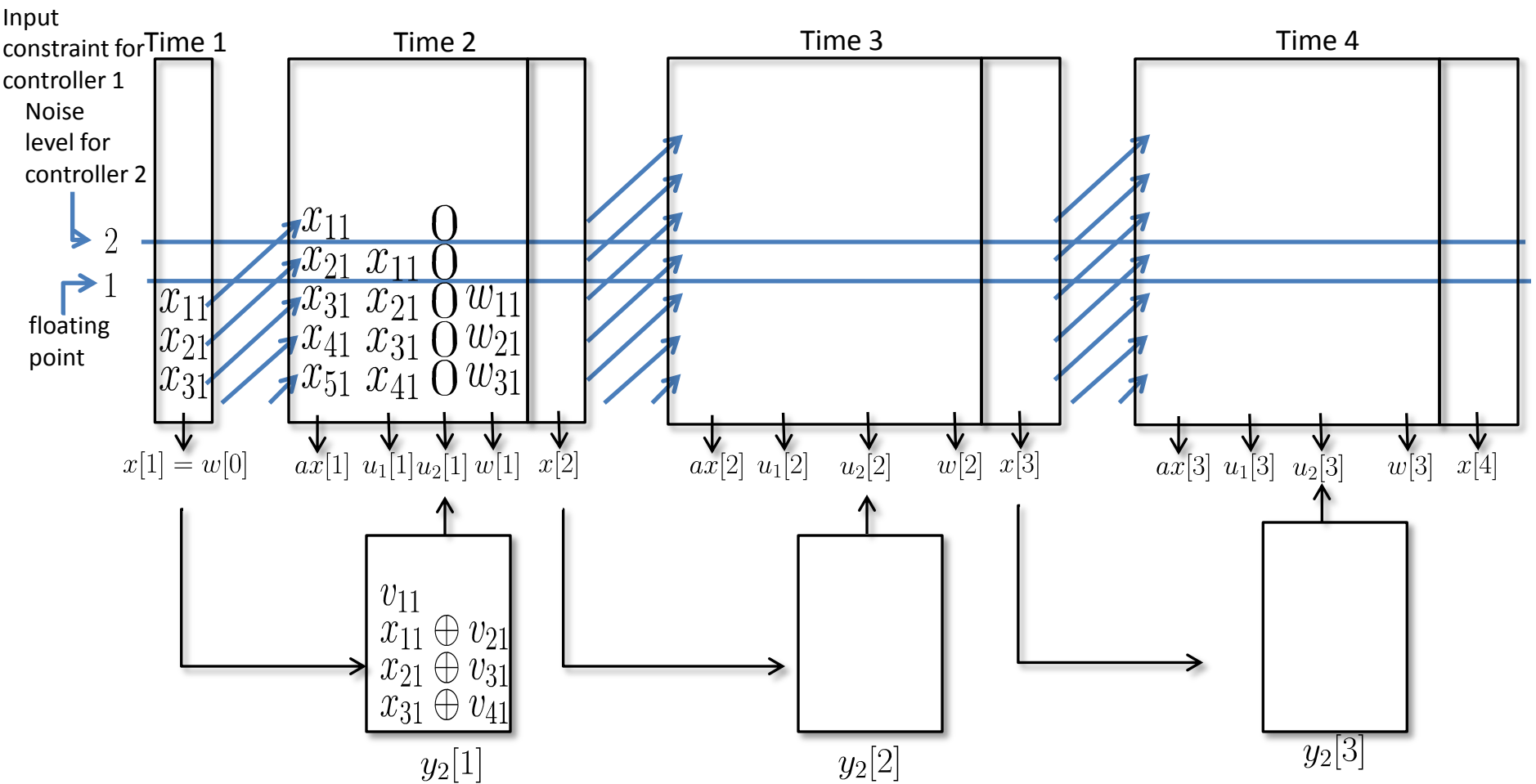
$$y_1[n] = x[n]$$

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$$w[n] \sim \mathcal{N}(0, 1)$$

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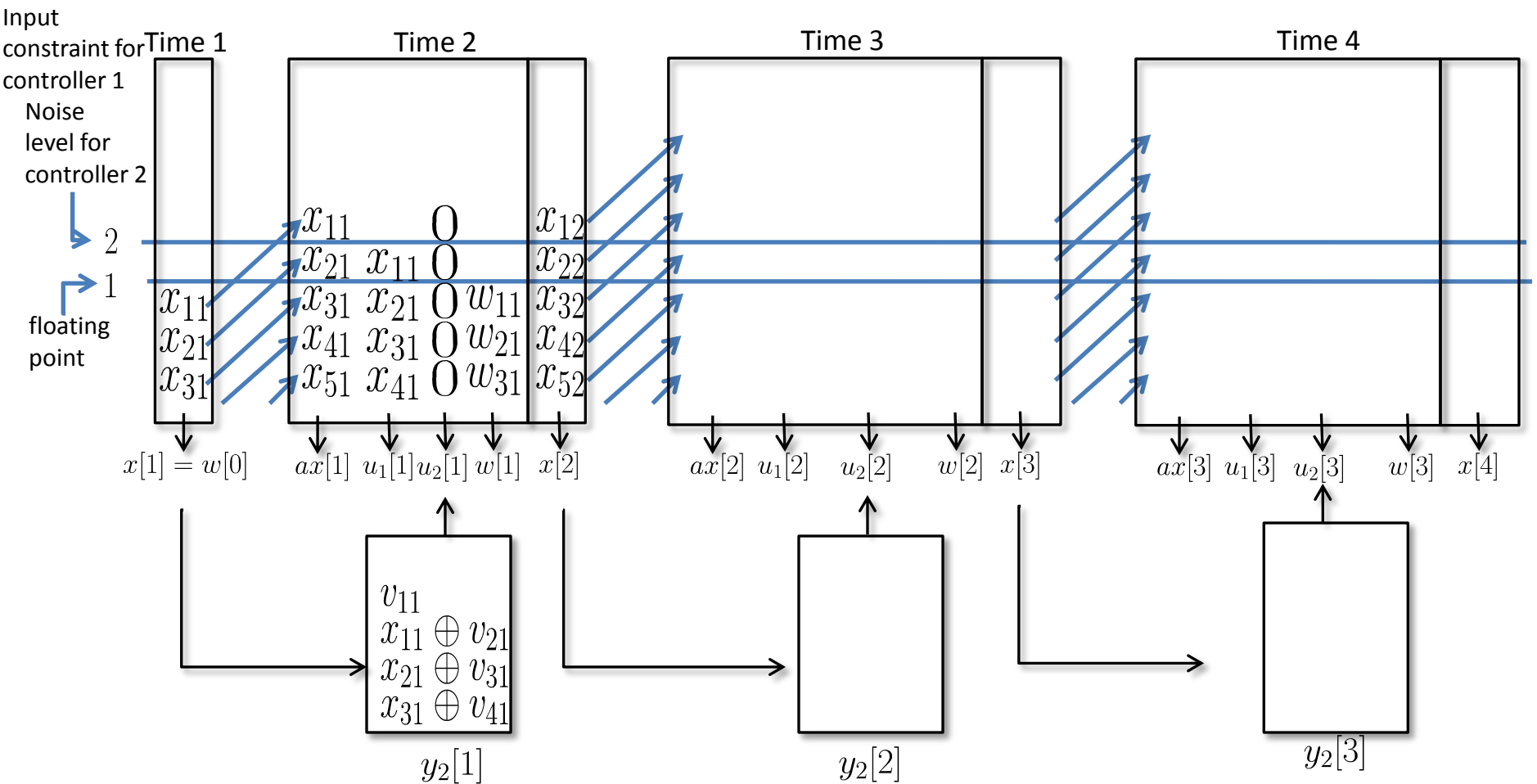
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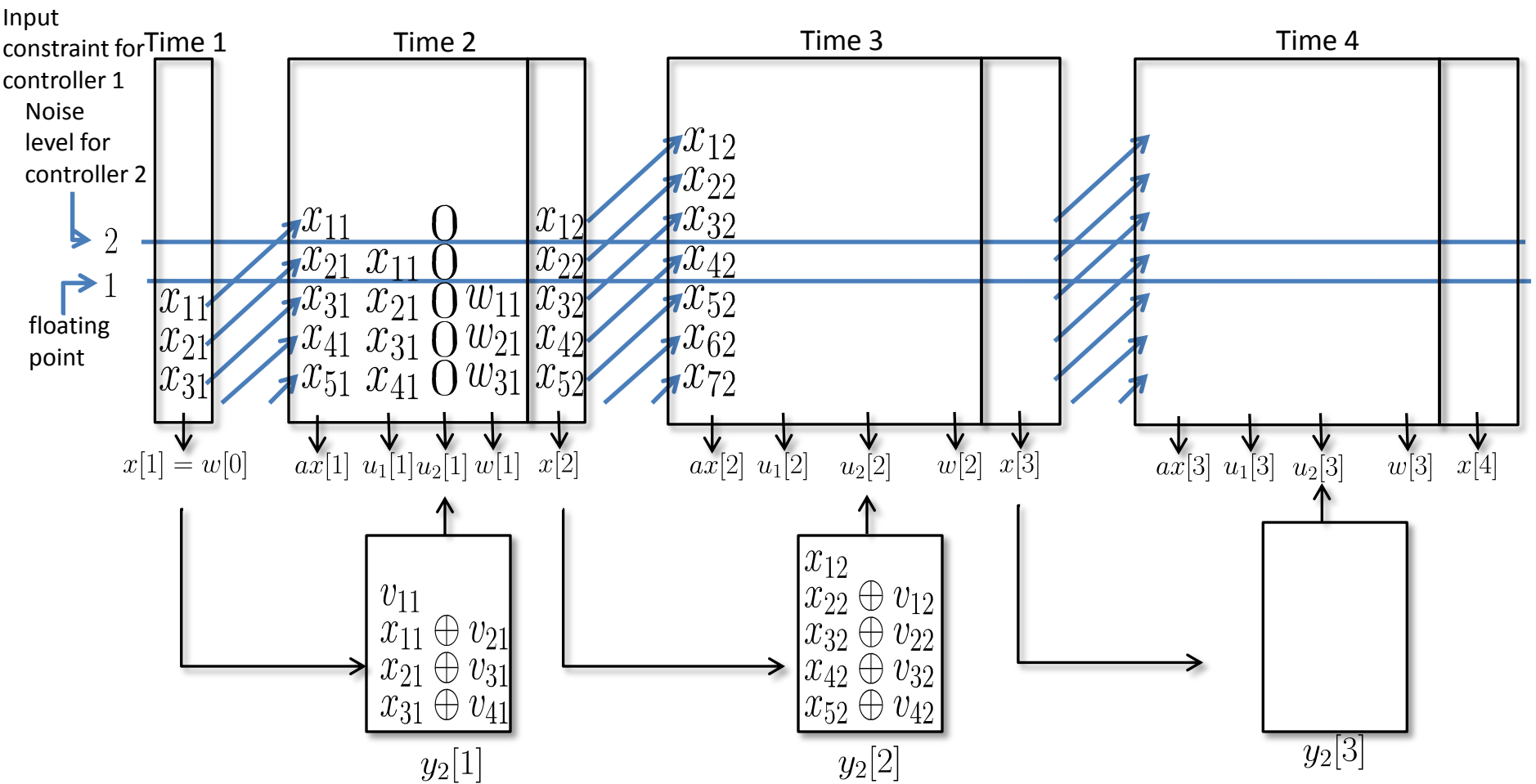
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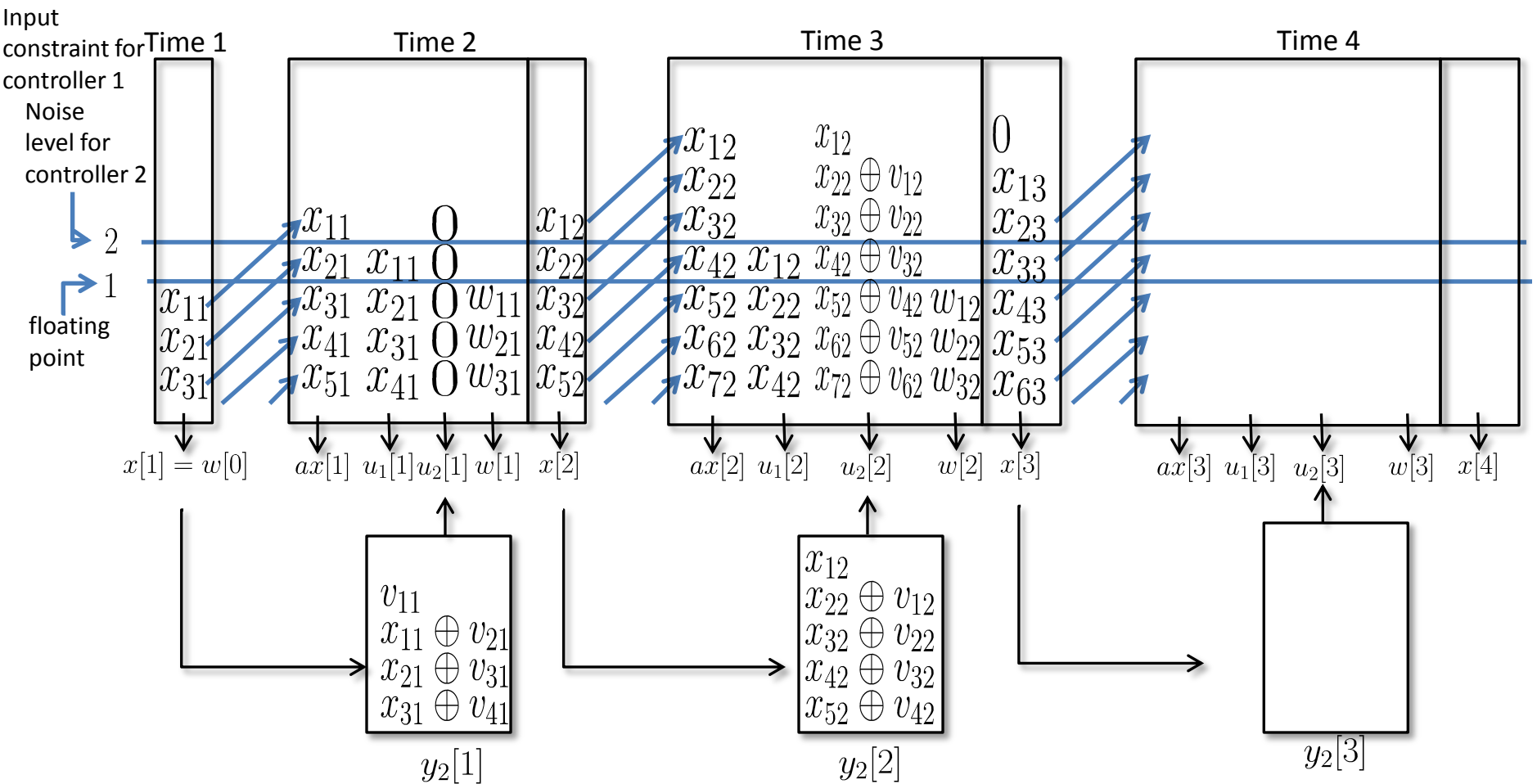
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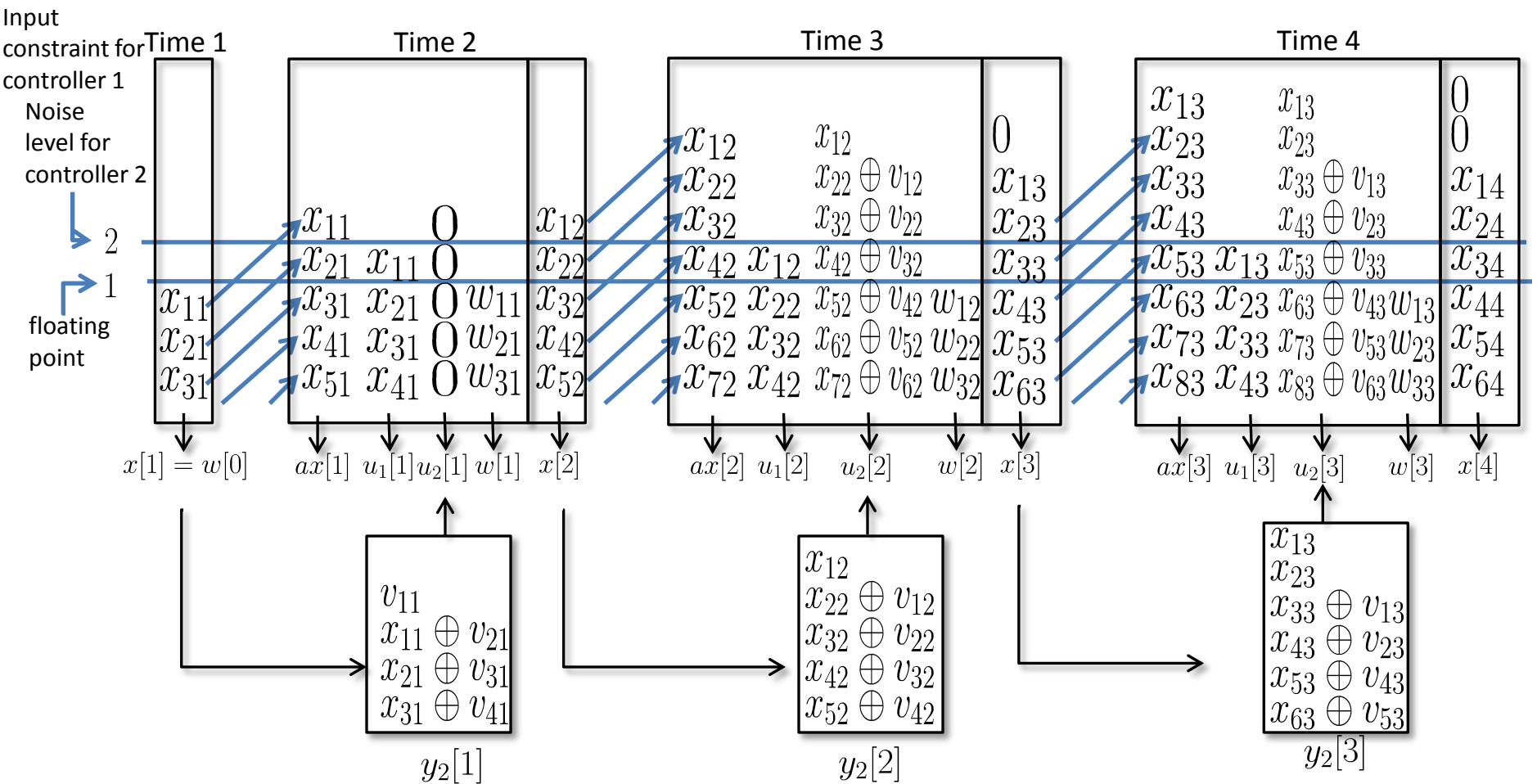
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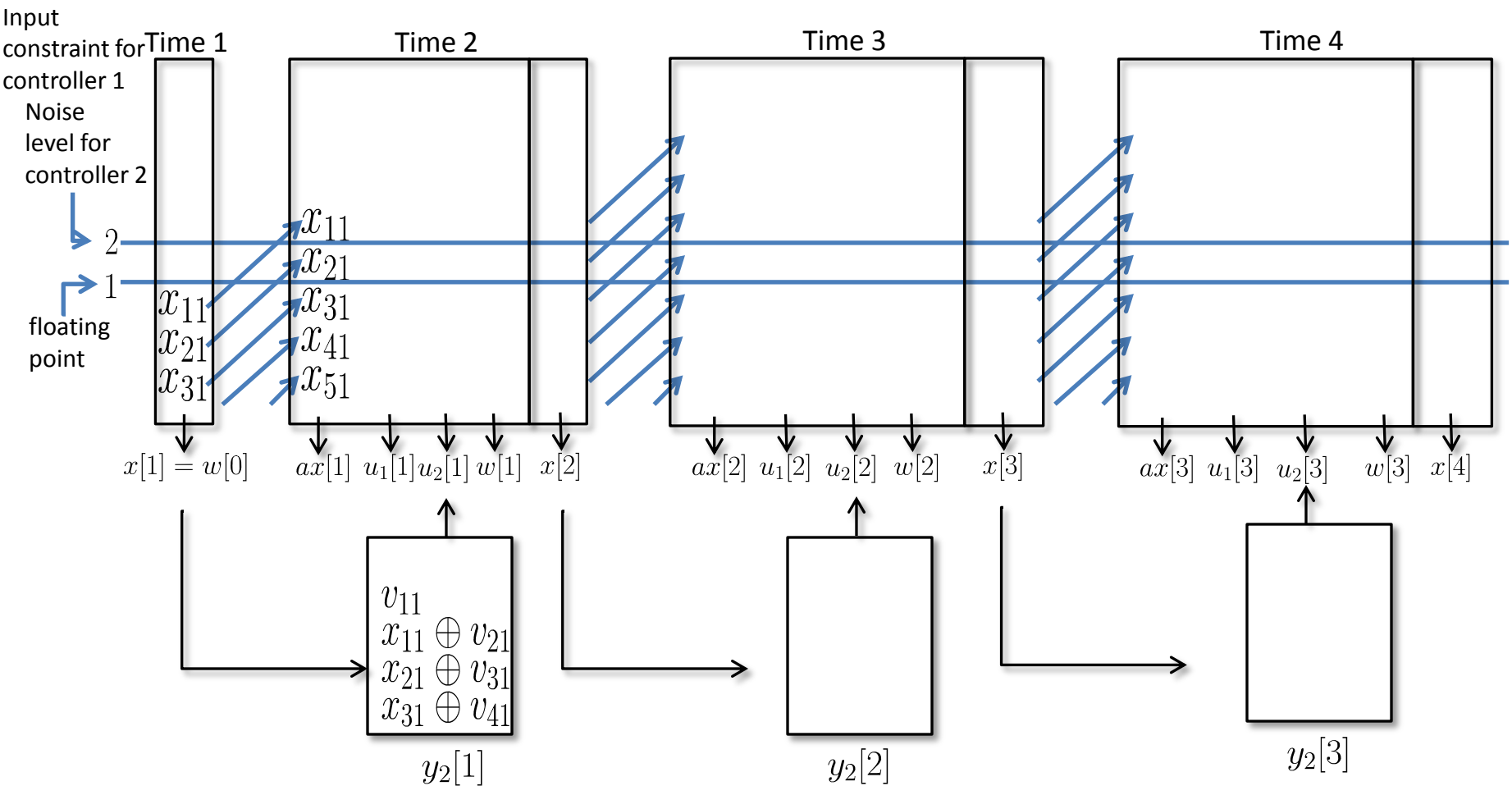
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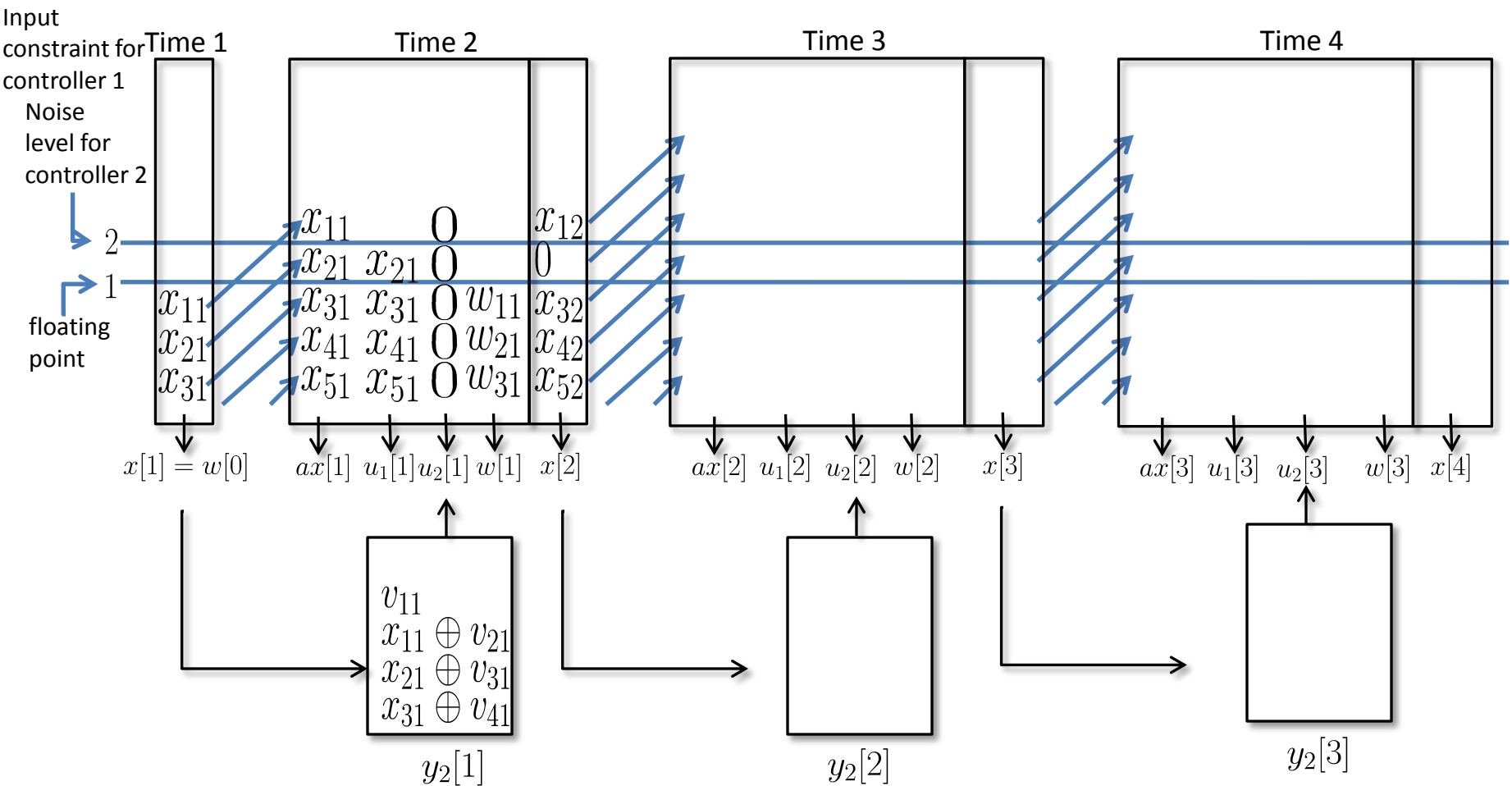
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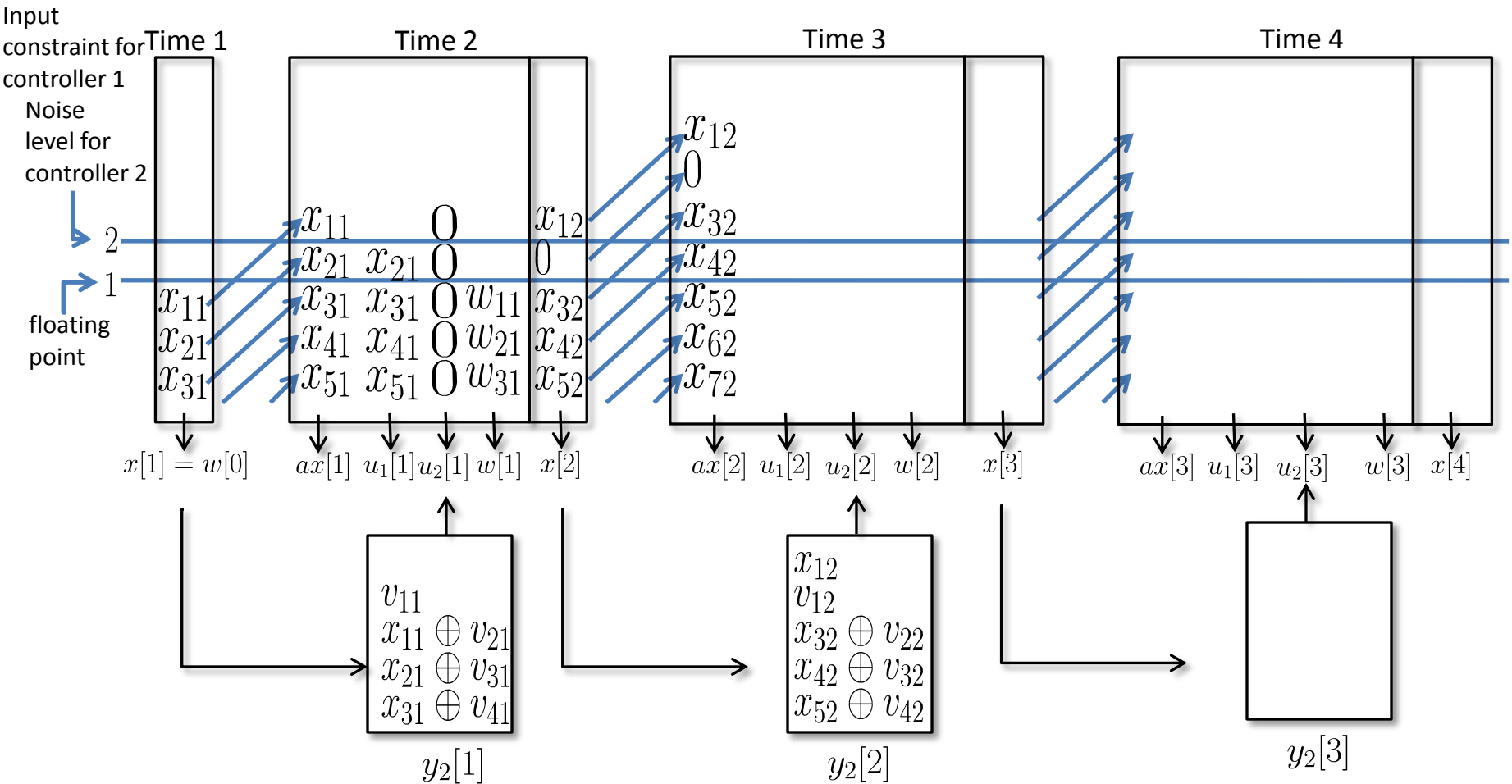
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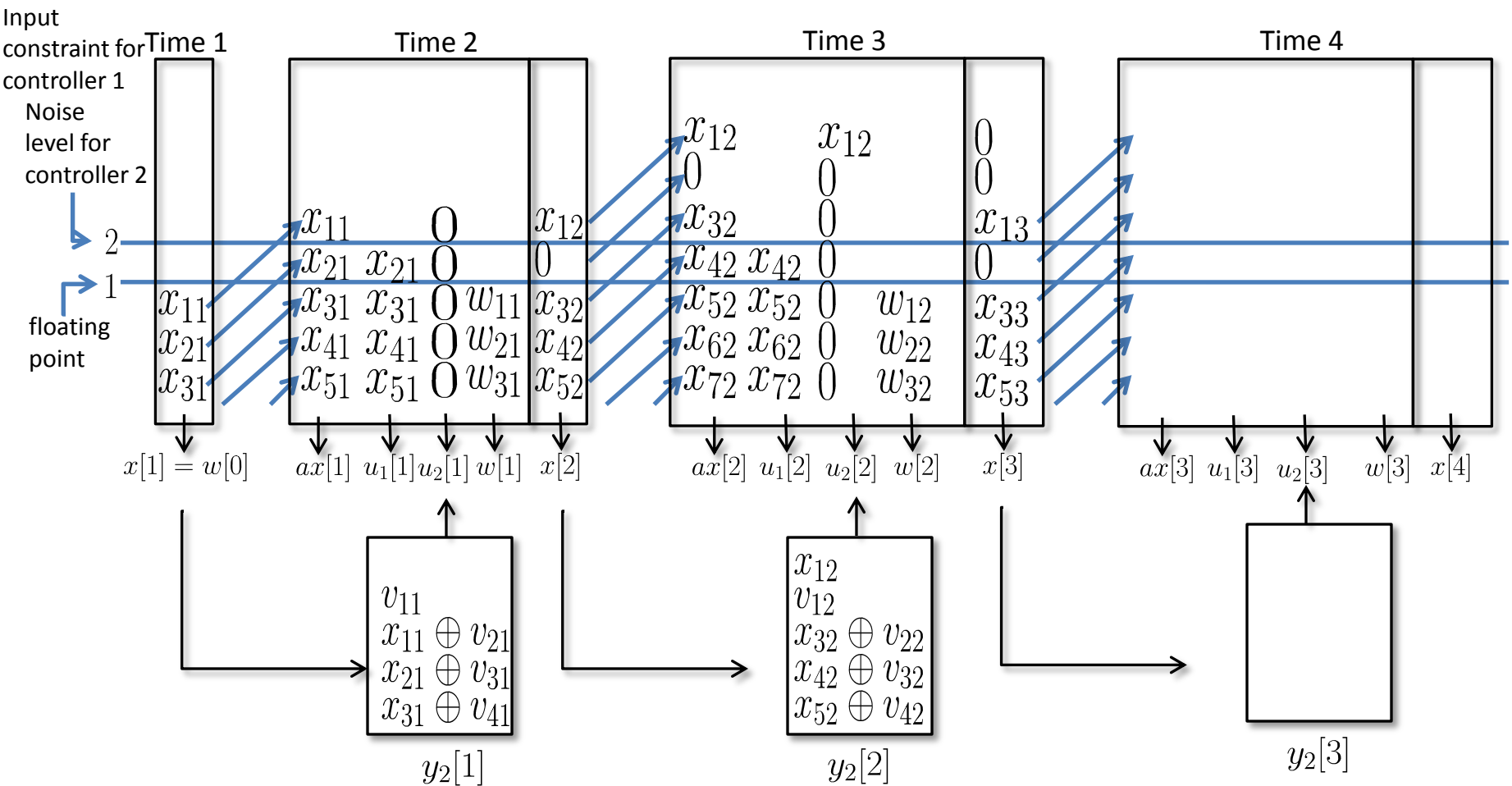
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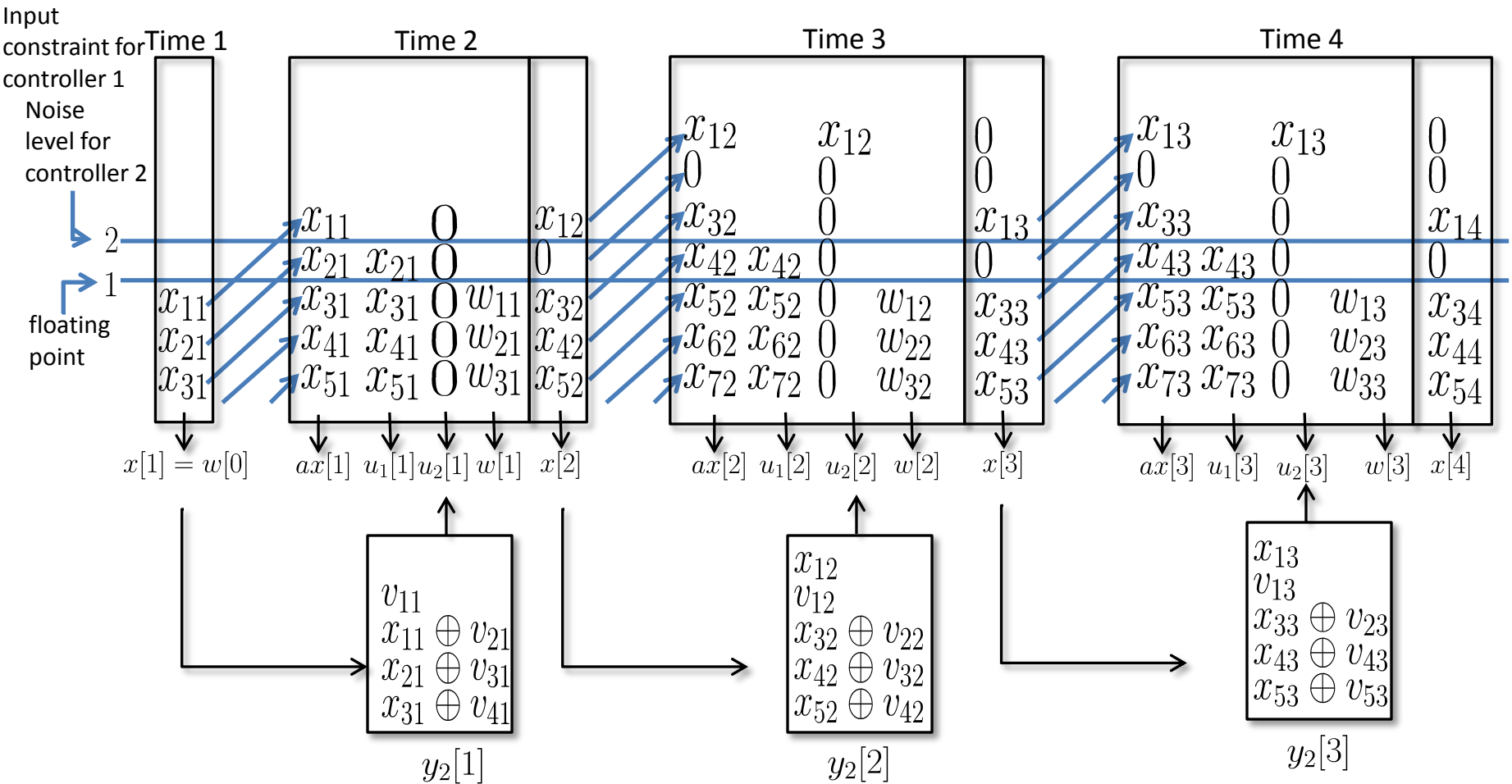
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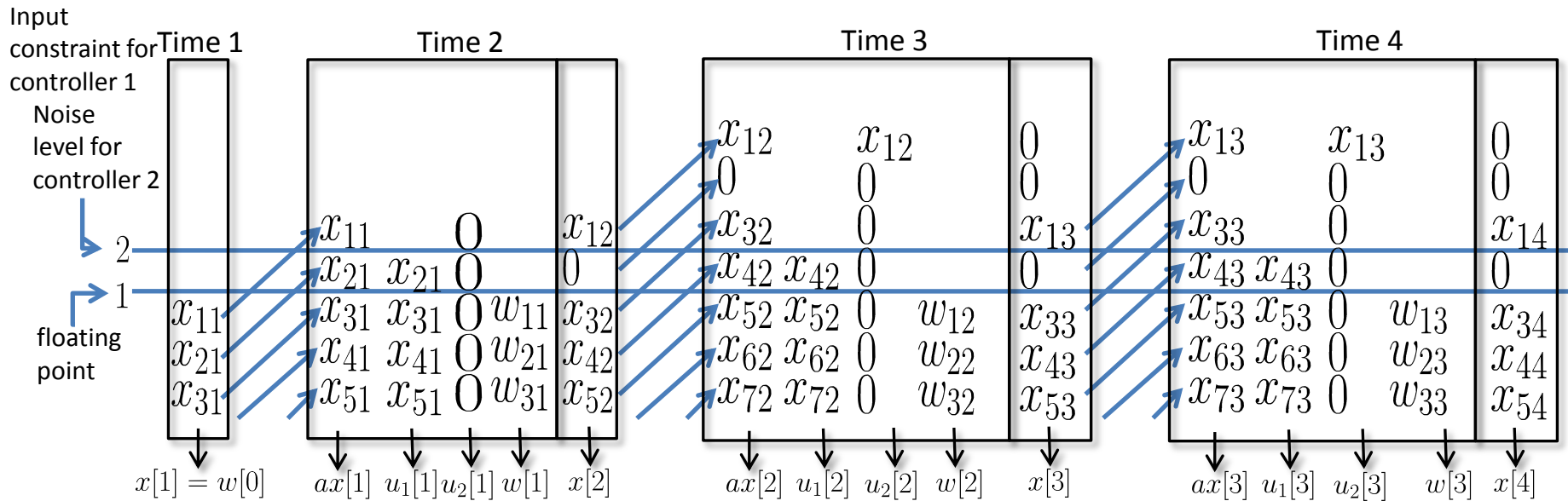
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Nonlinear Controller



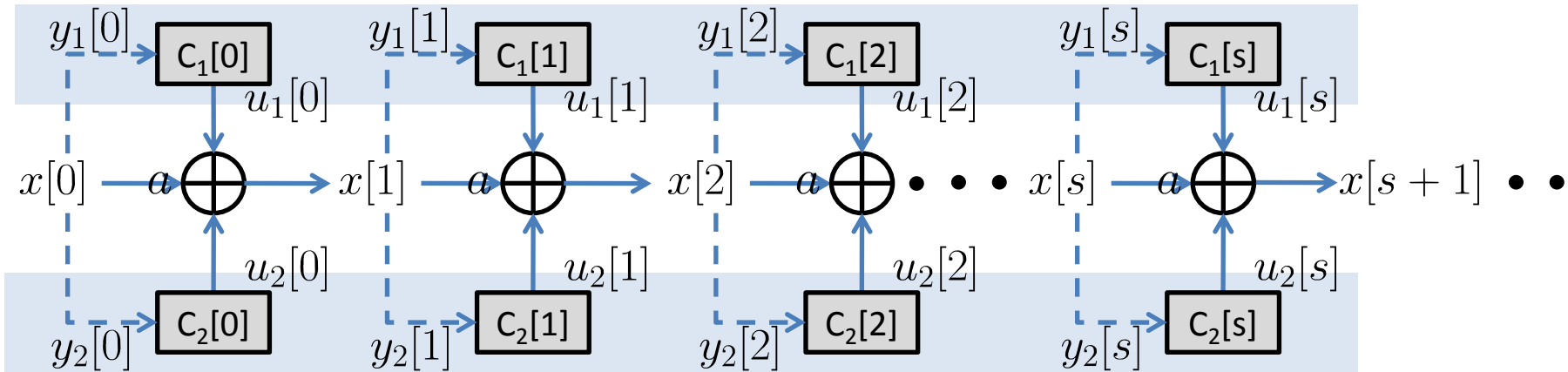
The corresponding scheme for the Reals

1-Stage Signaling Strategy $L_{sig,1}$

$$u_1[n] = -aR_d(y_1[n])$$

$$u_2[n] = -a(Q_{ad}(y_2[n] - R_{ad}(u_2[n-1])) + R_{ad}(u_2[n-1]))$$

Approximately Optimal Strategy

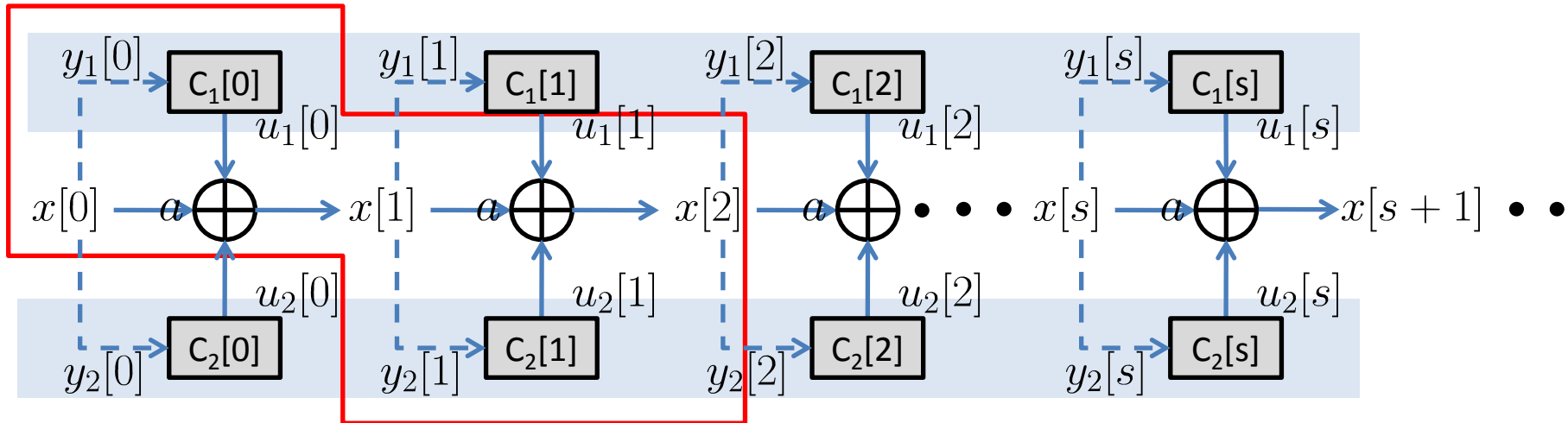


$$x[n + 1] = ax[n] + u_1[n] + u_2[n] + w[n]$$

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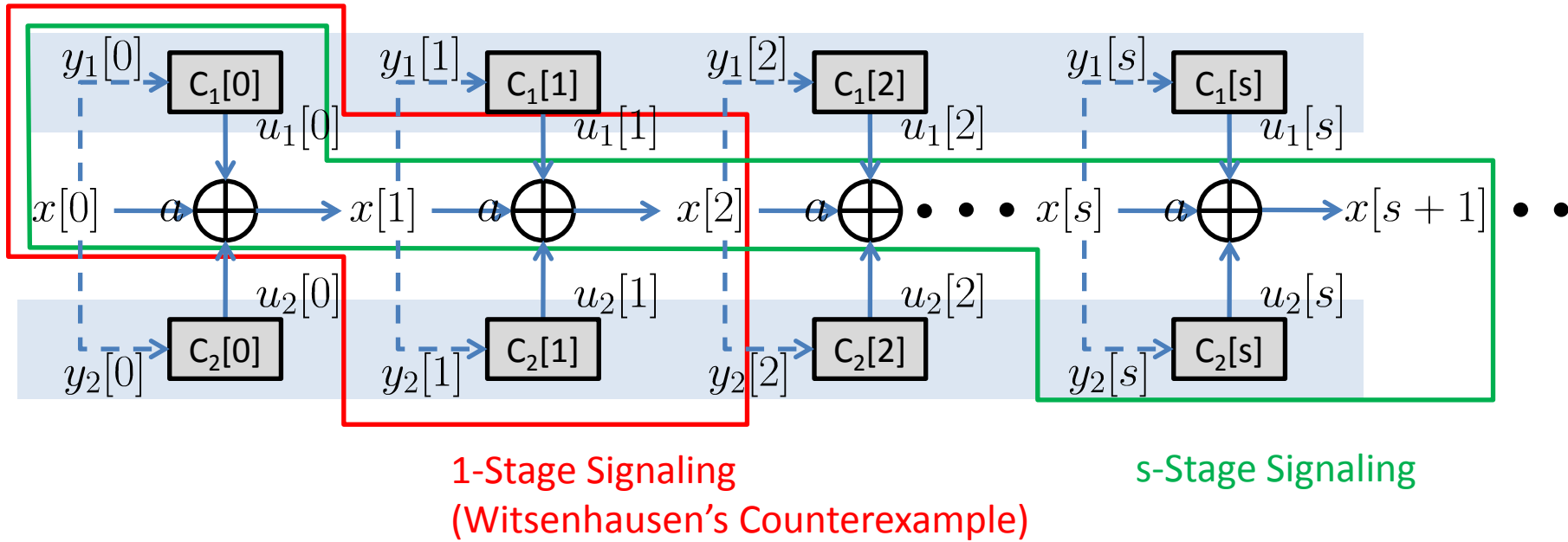
1-Stage Signaling
(Witsenhausen's Counterexample)

1-Stage Signaling Strategy $L_{sig,1}$

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Approximately Optimal Strategy



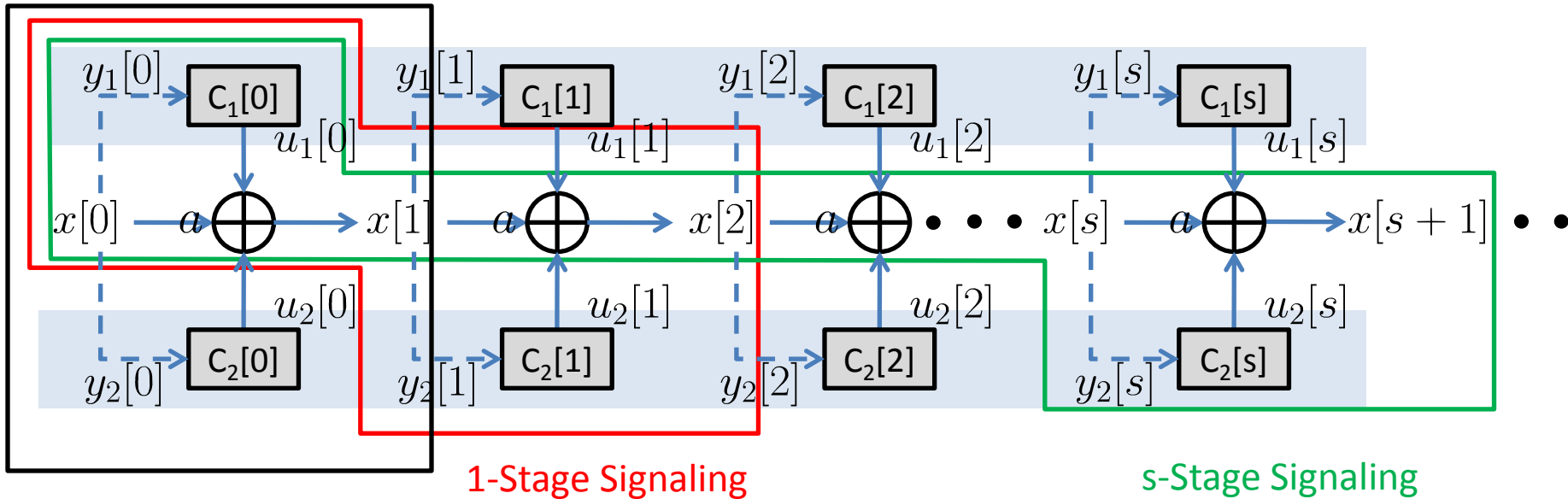
s -Stage Signaling Strategy $L_{sig,s}$

$$u_1[n] = -aR_d(y_1[n])$$

$$u_2[n] = -a(Q_{a^s d}(y_2[n]) - R_{a^s d}(\sum_{1 \leq i \leq s} a^{i-1} u_2[n-i]))$$

$$+ R_{a^s d}(\sum_{1 \leq i \leq s} a^{i-1} u_2[n-i]))$$

Approximately Optimal Strategy



Radner's Problem

(0-Stage Signaling is impossible)

1-Stage Signaling
(Witsenhausen's Counterexample)

s-Stage Signaling

Linear Strategy L_{lin}

$$u_1[n] = 0$$

$$u_2[n] = -ay_2[n]$$

or

$$u_1[n] = -ay_1[n]$$

$$u_2[n] = 0$$

Approximately Optimal Strategy

$$x[n + 1] = ax[n] + u_1[n] + u_2[n] + w[n]$$

$$y_1[n] = x[n] + v_1[n]$$

$$y_2[n] = x[n] + v_2[n]$$

Theorem [Park and Sahai, 2012]

There exists $c < 6 \cdot 10^6$ such that for all $|a| \geq 4$, q , r_1 , r_2 , σ_w , σ_{v1} , σ_{v2}

$$\frac{\inf_{u_1, u_2 \in L_{lin} \cup \cup_s L_{sig, s}} \limsup_{N \rightarrow \infty} \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[qx^2[n] + r_1 u_1^2[n] + r_2 u_2^2[n]]}{\inf_{u_1, u_2} \limsup_{N \rightarrow \infty} \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[qx^2[n] + r_1 u_1^2[n] + r_2 u_2^2[n]]} \leq c$$

Slow Dynamics Case (When $|a| < 4$)

- SNR(Signal-to-Noise Ratio) for implicit communication between two controller is **bounded** by $|a|$.
- **Single Controller Linear Strategy** is optimal within a constant ratio.
- Unlike Fast Dynamics Case, we need **Kalman** filtering estimator.

Single Controller Optimal Strategy

$$x[n + 1] = ax[n] + u[n] + w[n]$$

$$y[n] = x[n] + v[n]$$

Average Cost:

$$\inf_u \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[qx^2[n] + ru^2[n]]$$

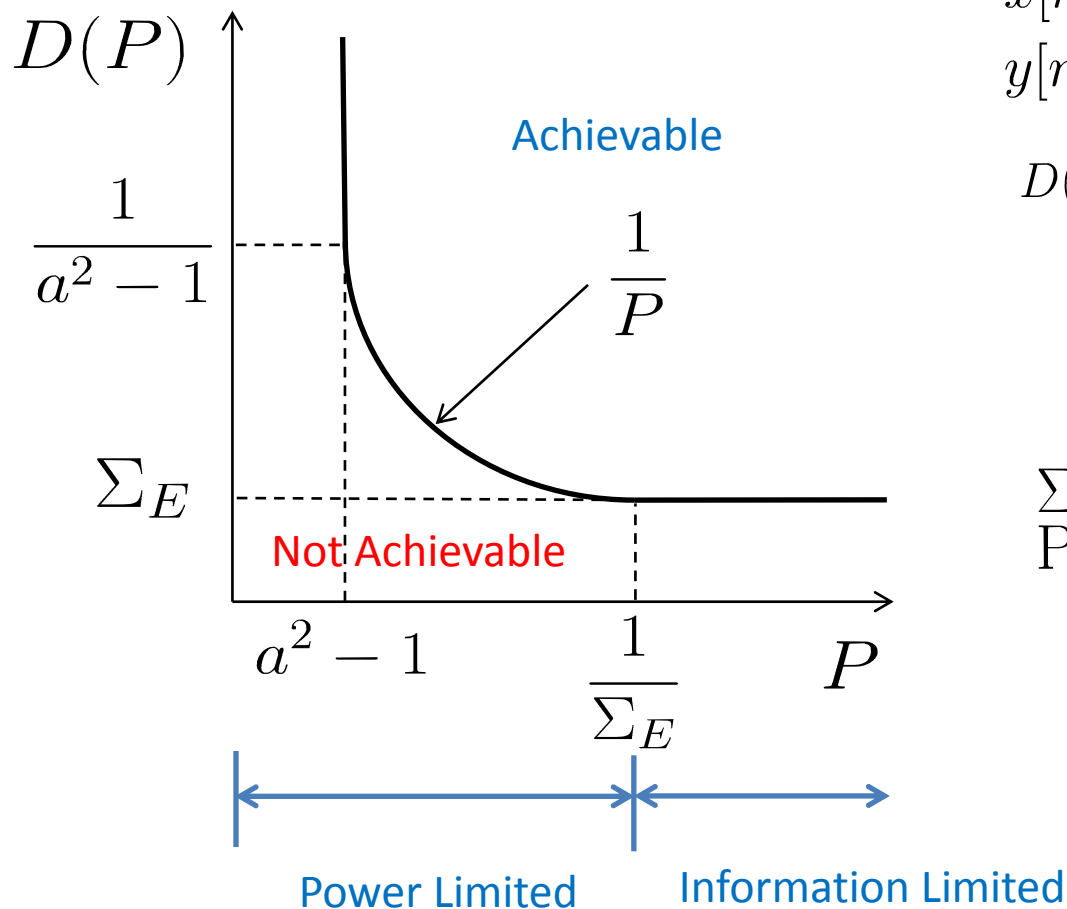
Power-Distortion Tradeoff:

$$D(P) = \inf_u \frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[x^2[n]]$$

s.t. $\frac{1}{N} \sum_{0 \leq n < N} \mathbb{E}[u^2[n]] \leq P$

Power-Distortion Tradeoff: When $1 < |a| < 4$

Conceptual Picture of the Tradeoff Curve



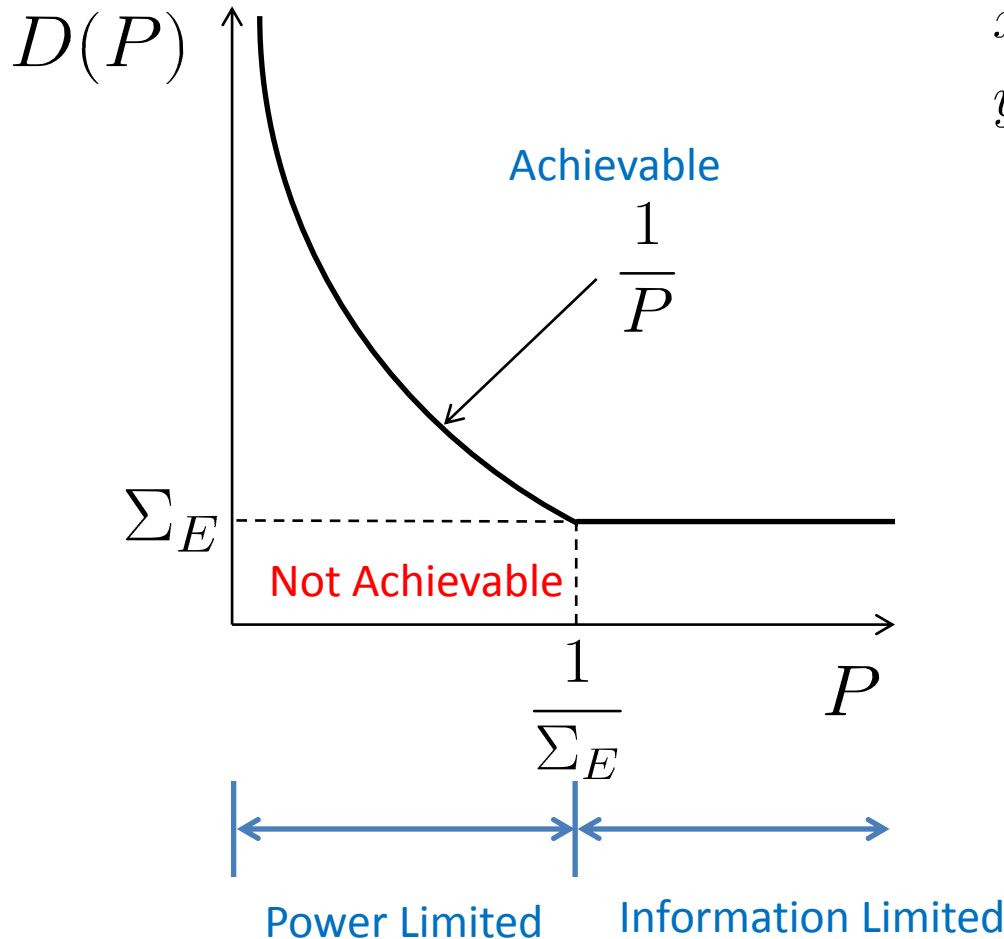
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Σ_E : Kalman Filter Performance

Power-Distortion Tradeoff: When $|a|=1$

Conceptual Picture of the Tradeoff Curve



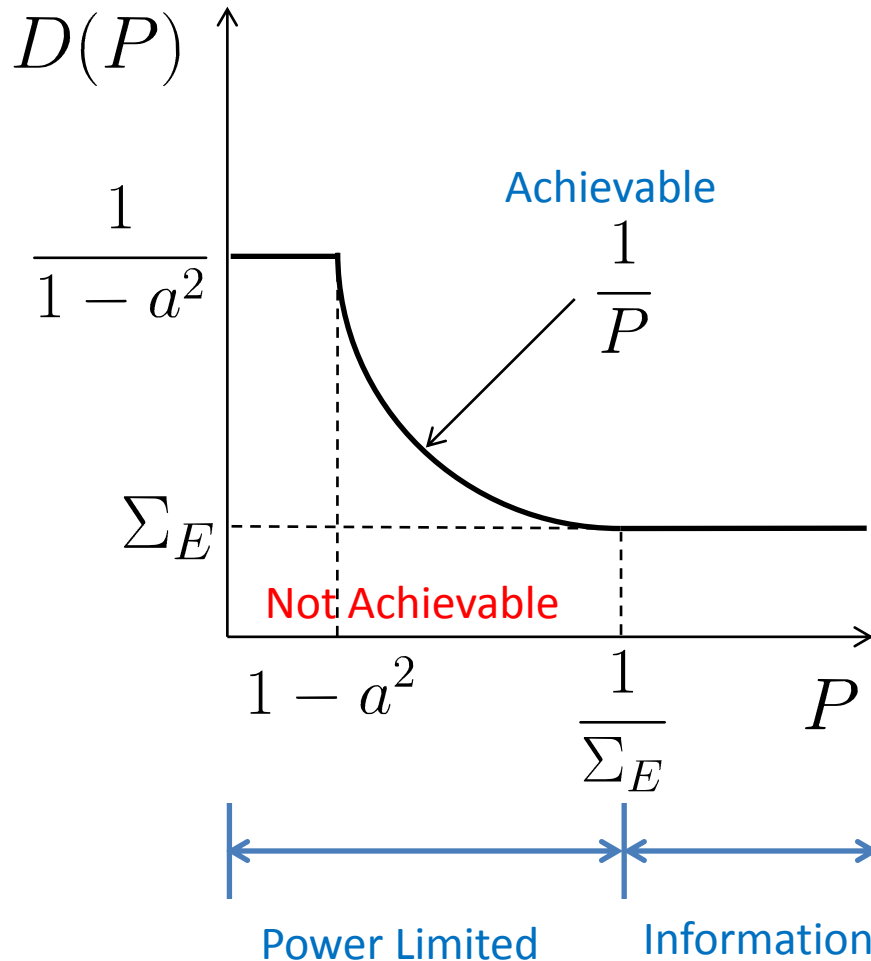
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Σ_E : Kalman Filter Performance

Power-Distortion Tradeoff: When $|a| < 1$

Conceptual Picture of the Tradeoff Curve



$$x[n+1] = ax[n] + u[n] + w[n]$$
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Σ_E : Kalman Filter Performance

Approximately Optimal Strategy

Linear Strategy L_{lin}

$$\begin{array}{l} u_1[n] = 0 \\ u_2[n] = -k\mathbb{E}[x[n]|y_2^n] \end{array} \quad \text{or} \quad \begin{array}{l} u_1[n] = -k\mathbb{E}[x[n]|y_1^n] \\ u_2[n] = 0 \end{array}$$

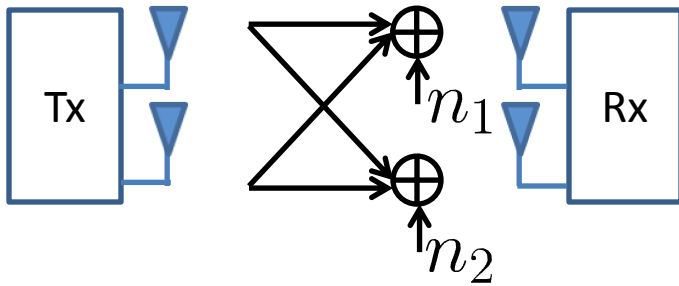
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Conclusion

MIMO Communication Problem



Linear Super-position of Signals, Gaussian Disturbance

Divide Cases:

(1) High-SNR (Signal-to-Noise Ratio)

- d.o.f. gain (rank of signal) is important
- Rank maximization scheme

(1) Low-SNR (Signal-to-Noise Ratio)

- Beam-forming gain (power of signal) is important
- Maximum-Ratio combining scheme

Decentralized Scalar LQG Problem

$$x[n + 1] = ax[n] + u_1[n] + u_2[n] + w[n]$$

$$y_1[n] = x[n] + v_1[n]$$

$$y_2[n] = x[n] + v_2[n]$$

Divide Cases:

How we get the Information for control?

(1) Fast Dynamics

- Implicit Communication from the other controller

(2) Slow Dynamics

- Kalman filtering gain is crucial

Conclusion

Centralized LQG Problem

$$\begin{aligned}x[n + 1] &= Ax[n] + Bu[n] + w[n] \\y[n] &= Cx[n] + v[n]\end{aligned}$$

- Finite-Dimensional Solution
- Estimation-Control Separation

Decentralized Scalar LQG Problem

$$\begin{aligned}x[n + 1] &= ax[n] + u_1[n] + u_2[n] + w[n] \\y_1[n] &= x[n] + v_1[n] \\y_2[n] &= x[n] + v_2[n]\end{aligned}$$

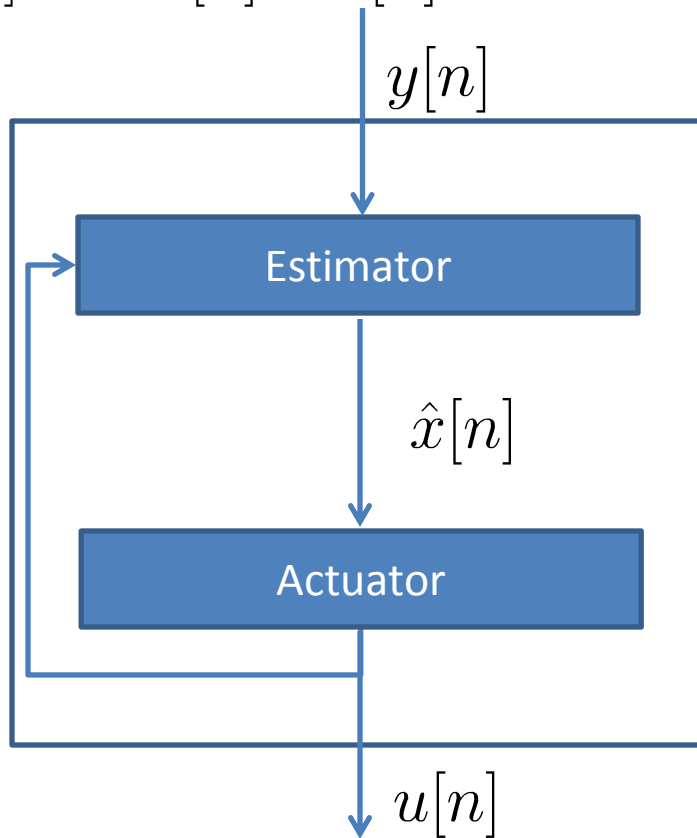
- Finite-Dimensional Approximated Solution
- Estimation-Control Separation and Implicit Communication Strategy

Conjecture

Centralized LQG Problem

$$x[n+1] = Ax[n] + Bu[n] + w[n]$$

$$y[n] = Cx[n] + v[n]$$



Estimation-Control Separation

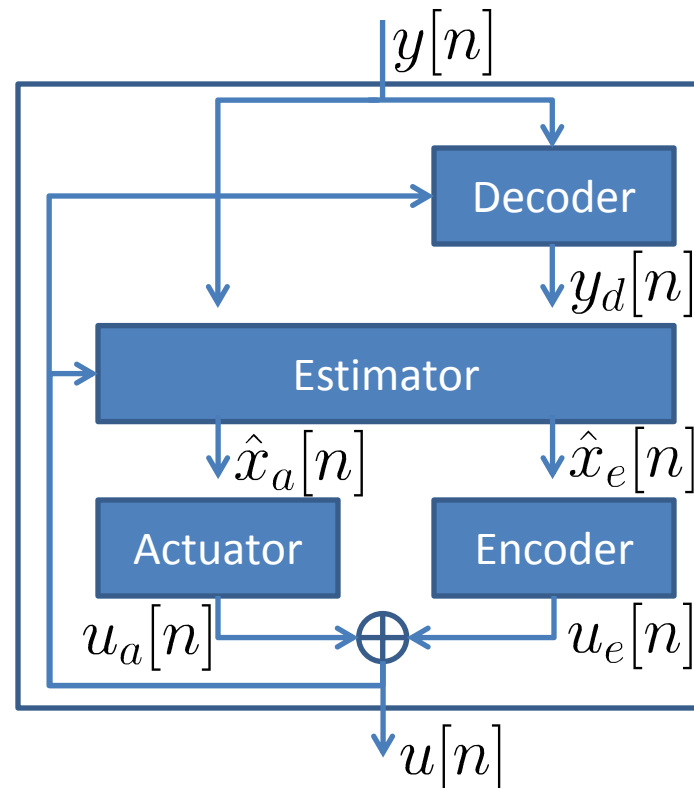
Decentralized LQG Problem

$$x[n+1] = Ax[n] + B_1u_1[n] + \dots + B_mu_m[n] + w[n]$$

$$y_1[n] = C_1x[n] + v_1[n]$$

⋮

$$y_m[n] = C_mx[n] + v_m[n]$$



Communication-Estimation-Control Separation

- Thank you